

SR80 Series Audio AI MCUs

Electrical Specification



Description

The Synaptics® Astra™ SR80 Series of Audio AI MCUs deliver industry-leading performance with ultra-low-power consumption, enabling the next generation of AI-enhanced, voice-driven, consumer and enterprise-grade IoT devices.

The AI MCUs combine an Arm® Cortex®-M33 MCU, Synaptics DSP cores, a Synaptics NPU, a Tensilica® HiFi 5 DSP, a built-in CODEC, high-speed USB, a rich set of I/O interfaces and advanced security features into a powerful platform designed for state-of-the-art wired and true-wireless (TWS) headsets, earbuds, voice-enabled smart-home devices, soundbars, and conferencing systems.

Turnkey AI-accelerated algorithms such as environmental noise cancellation (ENC) designed to exceed Microsoft® Teams™ version 5 Premium requirements, hybrid active noise cancellation (ANC), and speech and sound recognition, deliver natural, responsive voice experiences and human-machine interactions in real-world environments.

Open software and design collateral including the Astra SDK, audio performance tuning tools, reference designs, and evaluation hardware are supported on the SR80 Series to accelerate product development.

The SR80 Series gives product designers signal processing capability that far exceeds what is available today. An advanced **multi-core** audio processing engine provides a wealth of DSP resources to support advanced features. The NPU enables advanced AI features and support for TensorFlow™ Lite for Microcontrollers.

Additionally, a Tensilica HiFi 5 DSP, and an Arm® M33 processor complement the core signal processing engines. The Tensilica HiFi 5 DSP supports an Open-DSP platform for third party and customer application development. The Arm M33 core provides control and sensor fusion functions. The SR80 Series provides secure boot, secure firmware updates, secure debugging, user, and sensor data protection by utilizing integrated device identification, authentication, secure memory partitioning, and encryption functionality.

The integrated audio codec provides high performance, stereo (two) digital-to-analog converters (DACs), four analog-to-digital converters (ADCs), and eight digital microphones (DMIC) inputs.

High-performance class-AB/G headphone amplifiers complement the high dynamic range DACs for high fidelity playback (110 dB dynamic range).

Advanced continuous time delta-sigma ADCs paired with cap-less differential input amplifiers offer 106 dB of dynamic range for analog mic inputs for demanding applications.

The SR80 Series interfaces include USB 2.0 high-speed interface, three I2S, I3C host/target, two I²C hosts, one I²C target, three UART, two SPI host, one QSPI host/SPI target, PWM LED, 31 GPIOs, one 10-bit Monitor ADC, and JTAG debug.

A comprehensive set of software development and tuning tools, reference designs, and evaluation hardware is available to guide product development.

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1. Block Diagram

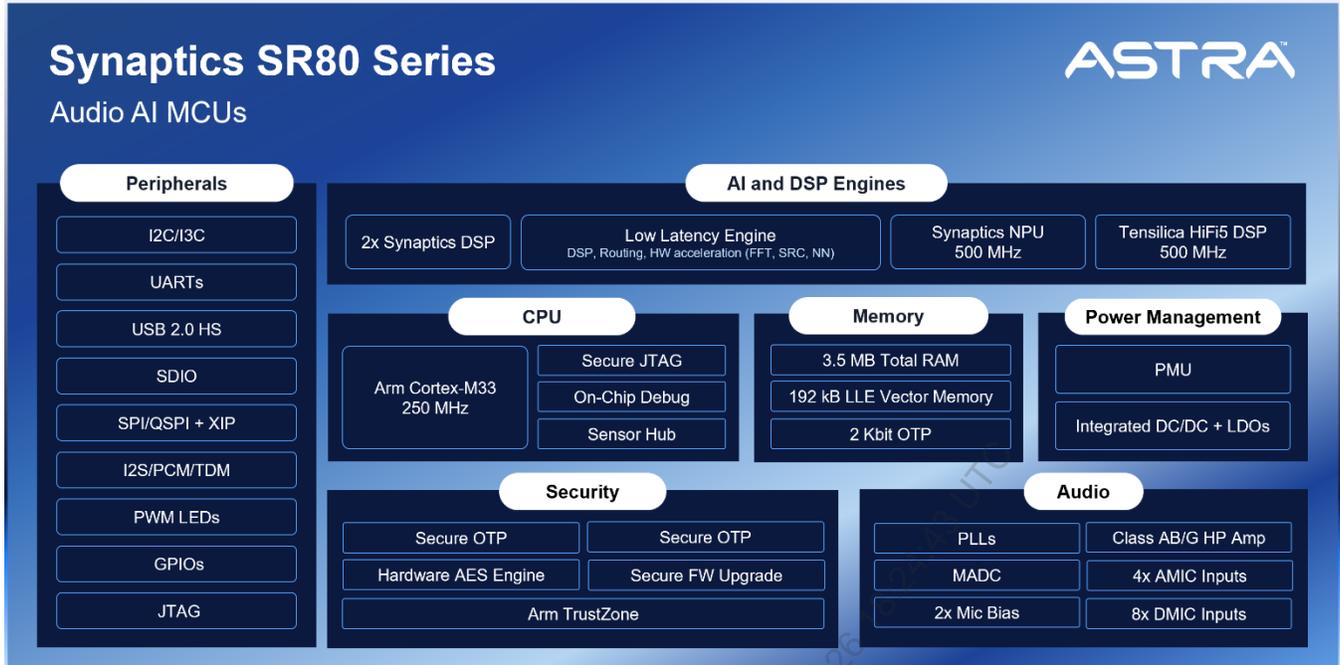


Figure 1. SR80 Series block diagram

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2. Features

- **Processors:** Arm Cortex–M33 CPU, HiFi 5 DSP, Synaptics dual CAPE 2 DSP, Synaptics LLE for ANC, and Synaptics NPU
- **Processing Subsystems**
 - Arm Cortex–M33 CPU: up to 250 MHz
 - HiFi 5 DSP (Tensilica®/Cadence®, 32-bit audio centric): up to 500 MHz
 - Synaptics low–power DSP subsystem
 - 2× CAPE 2 DSPs (32-bit IEEE floating–point + fixed–point)
 - Low Latency Engine (LLE) for ANC
 - up to 250 MHz
 - Synaptics NPU: up to 500 MHz
- **Memory**
 - 3.5 MB total memory
- **Security**
 - Hardware security features
 - Arm® TrustZone®
 - Device Identification and authentication
 - Secure / non–secure memory partitioning
 - Secure boot, firmware updates, and debugging
 - Hardware AES engine
 - OTP for key storage
 - Advanced encryption standard (AES) for encryption, decryption
- **Audio Processing**
 - Active Noise Cancellation
 - Ultra–low latency
 - Flexible mixing of anti–noise, ambient fusion, and playback signals
 - Support for multiple form factors/designs
 - Easy to use tuning tool
 - Supervisory controls
 - User adaptive
 - Environment adaptive
 - Voice Processing
 - Voice capture noise reduction optimized for all major electronic devices (High–Performance doorbells, intercoms, security cameras & panels, glass–break & other "Listeners")
 - High performance, low–power audio
 - Stereo 110 dBA dynamic range class–AB/G headphone amplifiers
 - Four 105 dBA differential microphone inputs
- **Power Management / Charging Interfaces**
 - One–wire PLC Communication

- **Connectivity Interfaces**
 - 3× Inter-IC Sound (I2S) / pulse-code modulation (PCM) / time division multiplexed (TDM) interfaces
 - USB 2.0 high-speed interface
 - One SD/SDIO-SDR50 support up to 100 MHz
 - One QSPI host interface (supports program execution from external flash memory), SPI target
 - Two Serial Peripheral Interface (SPI) hosts
 - Three I2C: two host interfaces and one target interface
 - I3C® interface supports host or target operation (MIPI improved inter-integrated circuit bus with a two-wire, sensor interface)
- **Debug**
 - Joint Test Action Group (JTAG) for debug
 - Three (3) Universal asynchronous receiver/transmitter (UART) for debug
- **Packaging**
 - 76-balls WLCSP, 0.35 mm pitch, 3.082 mm x 3.20 mm
 - 76-balls LGA, 6.00 mm x 6.50 mm

Table 1. SR80 Series Product Families Feature Summary

| PRODUCT FAMILY | SR82 | SR85 | SR88 |
|--------------------------------|--|---|--|
| FEATURES |  |  |  |
| MCU-Cortex-M33, DSP-CAPE2, LLE | Yes | Yes | Yes |
| HiFi5 | No | Yes | Yes |
| NPU | No | No | Yes |
| Grade & Package | | | |
| Package | WLCS-76 / LGA-76 | | |
| Consumer | Yes (TA: 0 to +70 ° C / TJ: -40 to +85 ° C) | | |

2.1. Overview

The SR80 Series device combines a low-power, secure, Arm Cortex-M33 32-bit microcontroller unit (MCU), an industry standard Tensilica® HiFi 5 32-bit fixed point DSP, a Synaptics low-power DSP subsystem, Synaptics NPU, industry standard interfaces, generous memory, and advanced algorithms into a powerful platform for advanced noise cancelling devices.

2.1.1. Analog Subsystem

The analog subsystem provides low-power, high-performance ADCs, DACs, amplifiers, and other functions.

Microphone path:

- Four ADCs for analog inputs
- Dynamic Range: 105 dB A-wt
- THD+N: -90 dB @ -1 dBFS
- Differential cap-less inputs
- Analog PGA with 0, 6, 12 dB gain
- Additional digital gain (boost)
- Two microphone bias voltage outputs, up to 3.0 V, 5 mA current capability
 - Independently programmable (100 mV steps)
 - Bypass mode outputs 1.8 V mic bias
- Four PDM interfaces for up to eight digital microphones

Analog playback Path:

- Two DACs
- High-performance stereo Class-AB/G differential headphone amplifier
- 30 mW/32 Ohms
- Dynamic Range: 110 dB A-wt
- THD+N: -95 dB @ -1 dBFS

2.1.2. Arm M33 CPU Subsystem

A secure, single core, 32-bit Arm M33 based SSE200 sub-system provides system and sensor management and other functions.

Hardware features:

- Arm Cortex-M33 Processor, up to 250 MHz
 - DSP/single instruction, multiple data (SIMD)
 - Single precision floating-point unit (FPU)
- 128 KB SRAM
- 16 KB I-Cache
- 16 kB D-Cache
- 96 KB ROM
- 64 KB I-TCM
- 64 KB D-TCM
- 64-bit wide AXI bus matrix
- QSPI flash controller

- Built-in system security
 - Secure boot and Secure firmware upgrade
 - Device identification and authentication
 - Data security for local storage and data over the network
 - Secure JTAG debugging
 - AES-128/256: advanced encryption standard
- Timer/counter/watchdog
- On-chip debug

Software features:

- System control
- Boot-up host
- Firmware update
- Communication with Bluetooth® subsystem
- Sensor hub
- SDK for third party development

2.1.3. HiFi 5 DSP Subsystem

A dedicated Tensilica HiFi 5 DSP plus memory subsystem handles customer and third-party algorithms.

The HiFi 5 DSP high-performance core is part of the Tensilica HiFi DSP family for audio, voice, and speech, offering low-energy, high-performance processing for audio and voice algorithms.

Hardware features:

- Tensilica HiFi 5 processor, up to 500 MHz
- 32-bit fixed-point
- 1 MB local RAM, plus additional 0.5 MB shared RAM (total 1.5 MB available memory)
- TCM
 - 64 KB instruction
 - 128 KB data
- Cache
 - 32 KB instruction
 - 32 KB data
 - Prefetch
- On-chip debug (OCD)
- Configurable interrupt controller external to the core

Software features:

- Fully open platform for customer use (SDK available)
- FreeRTOS™
- Software API to the rest of the SR80 Series chip
- Control interface to applications running on other processors
- Control interface to hardware configuration
- Configurable audio routing to/from other points in the SR80 Series system

2.1.4. Synaptics DSP Subsystem

The Synaptics DSP Subsystem includes dual CAPE 2 cores and a low-latency engine (LLE) for ANC.

Hardware features:

- Dual CAPE 2 DSPs, up to 250 MHz
- 32-bit IEEE floating-point and 32-bit fixed-point
- Dual multiply/add + dual load/store + dual index update, all in one cycle
- Two floating point MACs per cycle
- Two 32-bit, four 16-bit, or eight 8-bit fixed-point multiply-accumulate operations (MACs) per cycle
- Program cache
- Low latency engine for ANC
 - Extension to architecture
 - Complex vector arithmetic
 - Fast Fourier transform (FFT) acceleration
 - Bi-quad filter acceleration
 - Sample rate conversion (SRC) (power of 2) acceleration
 - 24 bit, 8 bit, and ternary neural network acceleration

Software features:

- Environmental noise cancellation (ENC)
- Active Noise Cancellation (ANC)
- Acoustic echo cancellation (AEC)
- Voice wake-up/voice trigger
- Sample rate conversion
- Equalizer (EQ)
- Dynamic range control
- Neural networks

2.1.5. Synaptics NPU

Synaptics easy to use and highly optimized AI Neural Network acceleration processor working seamlessly with HiFi5 DSP supporting TensorFlow Lite for Microcontrollers (TFLM), for advanced voice and audio applications.

- 500 MHz maximum core speed.
- 32 (8x8 and 16x8) MACs
- 75% average MAC usage
- 32@500 MHz GOPS
- 32 kB local buffer and flexible allocation of low-latency shared memory

2.1.6. Interfaces

A generous offering of general-purpose input/output (GPIO) and industry standard interfaces makes connecting to external memories, sensors, SoCs, buttons, and indicators easy.

- USB 2.0 high-speed interface
- Three I2S/PCM/TDM host/target interfaces
- Two I²C host interfaces
- One I²C target interface (or I3C host/target)
- JTAG debug interface
- Three UART interfaces
- One QSPI host/SPI target interface
- Two SPI host interfaces
- 32 GPIOs

2.1.7. Timers

The system has multiple counter/timers that provide periodic triggers, count incoming events, or provide watchdog capability for system robustness.

2.1.8. Measurement ADC

A 10-bit measurement ADC (MADC) can be used for multi-button sensing, or for general-purpose monitoring of system voltages.

2.1.9. Power Management Unit

A dedicated power management unit (PMU) manages the multiple power domains and primary power states within the SR80 Series. The PMU provides controls for firmware to further manage power states and generate interrupts from wake events.

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3. Pin Information

3.1. Pin Definitions

Table 2. Pin type and direction abbreviations

| Abbreviation | Definition |
|------------------|-----------------------|
| Type | |
| A | Analog |
| D | Digital |
| OD | Open Drain |
| Diff | Differential (analog) |
| P | Power |
| GND | Ground |
| Direction | |
| In | Input only |
| Out | Output only |
| In/Out | Input and output |

Table 3. Signal definitions

| Signal | Ball | Pad | Type, Direction | Description |
|--|------|-----|-----------------|----------------------------------|
| I2S/PCM Interfaces | | | | |
| I2SO_PCM_BCLK | E15 | 59 | D, IN/OUT | I2SO bit clock |
| I2SO_PCM_WS | C11 | 60 | D, IN/OUT | I2SO word select |
| I2SO_PCM_TX_DAT | B14 | 54 | D, OUT | I2SO TX data |
| I2SO_PCM_RX_DAT | B12 | 56 | D, IN | I2SO RX data |
| I2S1_RX_DAT | K16 | 39 | D, IN | I2S1 RX data |
| I2S1_WS | K14 | 41 | D, IN/OUT | I2S1 word select |
| I2S1_TX_DAT | F14 | 43 | D, OUT | I2S1 TX data |
| I2S1_BCLK | G15 | 47 | D, IN/OUT | I2S1 bit clock |
| I2S1_RX_DAT_b | J13 | 37 | D, IN | I2S1 RX data (alternate pin) |
| I2S1_TX_DAT_b | J15 | 38 | D, OUT | I2S1 TX data (alternate pin) |
| I2S2_BCLK | G11 | 29 | D, IN/OUT | I2S2 bit clock |
| I2S2_WS | J11 | 31 | D, IN/OUT | I2S2 word select |
| I2S2_TX_DAT | K12 | 33 | D, OUT | I2S2 TX data |
| I2S2_RX_DAT | K10 | 35 | D, IN | I2S2 RX data |
| I2S2_BCLK_b | H10 | 25 | D, IN/OUT | I2S2 bit clock (alternate pin) |
| I2S2_WS_b | J9 | 27 | D, IN/OUT | I2S2 word select (alternate pin) |
| I2S2_TX_DAT_b | G9 | 32 | D, OUT | I2S2 TX data (alternate pin) |
| I2S2_RX_DAT_b | H8 | 34 | D, IN | I2S2 RX data (alternate pin) |
| I²C Target, I²C Host/Target, I³C Host/Target | | | | |

| Signal | Ball | Pad | Type, Direction | Description |
|---------------------------------|------|-----|-----------------|---|
| I2C_SLV_SCL/I3C_SCL | D12 | 50 | OD, I/O | I2C target clock I3C host/target clock |
| I2C_SLV_SDA/I3C_SDA | C13 | 48 | OD, I/O | I2C target data I3C host/target data |
| I3C_SCL_b | J15 | 38 | OD, I/O | I3C host/target clock (alternate pin) |
| I3C_SDA_b | J13 | 37 | OD, I/O | I3C host/target data (alternate pin) |
| I2CO_MS_SCL | J15 | 38 | OD, I/O | I2CO host/target clock |
| I2CO_MS_SDA | J13 | 37 | OD, I/O | I2CO host/target data |
| I2C1_MS_SCL | H16 | 40 | OD, I/O | I2C1 host/target clock |
| I2C1_MS_SDA | H14 | 36 | OD, I/O | I2C1 host/target data |
| I2C1_MS_SCL_b | J7 | 28 | OD, I/O | I2C1 host/target clock (alternate pin) |
| I2C1_MS_SDA_b | K8 | 30 | OD, I/O | I2C1 host/target data (alternate pin) |
| UART | | | | |
| UART0_RX | D12 | 50 | D, IN | UART0 RX data in |
| UART0_TX | C13 | 48 | D, OUT | UART0 TX data out |
| UART0_CTS | K12 | 33 | D, IN | UART0 CTS flow control |
| UART0_RTS | K10 | 35 | D, OUT | UART0 RTS flow control |
| UART0_CTS_b | J7 | 28 | D, IN | UART0 CTS flow control (alternate pin) |
| UART0_RTS_b | K8 | 30 | D, OUT | UART0 RTS flow control (alternate pin) |
| UART0_RX_b | J11 | 31 | D, IN | UART0 RX data in (alternate pin) |
| UART0_TX_b | G11 | 29 | D, OUT | UART0 TX data out (alternate pin) |
| UART0_RX_c | J9 | 27 | D, IN | UART0 RX data in (alternate pin) |
| UART0_TX_c | H10 | 25 | D, OUT | UART0 TX data out (alternate pin) |
| UART1_RX | A15 | 58 | D, IN | UART1 RX data in |
| UART1_TX | D14 | 53 | D, OUT | UART1 TX data out |
| UART1_RX_b | J13 | 37 | D, IN | UART1 RX data in (alternate pin) |
| UART1_TX_b | J15 | 38 | D, OUT | UART1 TX data out (alternate pin) |
| UART2_RX | H14 | 36 | D, IN | UART2 RX data in |
| UART2_TX | H16 | 40 | D, OUT | UART2 TX data out |
| UART2_RX_b | J9 | 27 | D, IN | UART2 RX data in (alternate pin) |
| UART2_TX_b | B10 | 66 | D, OUT | UART2 TX data out (alternate pin) |
| QSPI Host, SPI Target, SPI Host | | | | |
| QSPI_CLK/SPI_SL_SCK | B16 | 57 | D, IN/OUT | QSPI CLK SPI target CLK |
| QSPI_CSO/SPI_SL_CS | E13 | 49 | D, IN/OUT | QSPI chip select SPI target chip select |
| QSPI_SI/QSPI_SIO0/SPI_SL_SDI | C15 | 55 | D, IN/OUT | QSPI SI/SIO0 SPI target SDI |
| QSPI_SO/QSPI_SIO1/SPI_SL_SD | D16 | 51 | D, IN/OUT | QSPI SO/SIO1 SPI target SDO |
| QSPI_SIO2 | D14 | 53 | D, IN/OUT | QSPI SIO2 |
| QSPI_SIO3 | A15 | 58 | D, IN/OUT | QSPI SIO3 |
| SPI_SL_SCK_b | D12 | 50 | D, IN | SPI target CLK (alternate pin) |

| Signal | Ball | Pad | Type, Direction | Description |
|----------------------------|------|-----|-----------------|--|
| SPI_SL_CS_b | K16 | 39 | D, IN | SPI target chip select (alternate pin) |
| SPI_SL_SDI_b | C13 | 48 | D, IN | SPI target SDI (alternate pin) |
| SPI_SL_SDO_b | F14 | 43 | D, OUT | SPI target SDO (alternate pin) |
| SPI1_MSTR_CLK | G11 | 29 | D, OUT | SPI1 host clock |
| SPI1_MSTR_CS | J11 | 31 | D, OUT | SPI1 host chip select |
| SPI1_MSTR_MOSI | K12 | 33 | D, OUT | SPI1 host data out |
| SPI1_MSTR_MISO | K10 | 35 | D, IN | SPI1 host data in |
| SPI2_MSTR_CLK | H10 | 25 | D, OUT | SPI2 host clock |
| SPI2_MSTR_CS | J9 | 27 | D, OUT | SPI2 host chip select |
| SPI2_MSTR_MOSI | G9 | 32 | D, OUT | SPI2 host data out |
| SPI2_MSTR_MISO | H8 | 34 | D, IN | SPI2 host data in |
| SDIO Interface | | | | |
| SD_CLK | H10 | 25 | D, IN | SDIO clock |
| SD_DAT0 | H8 | 34 | D, IN/OUT | SDIO data line 0 |
| SD_DAT1 | J7 | 28 | D, IN/OUT | SDIO data line 1 |
| SD_DAT2 | K8 | 30 | D, IN/OUT | SDIO data line 2 |
| SD_DAT3 | J9 | 27 | D, IN/OUT | SDIO data line 3 |
| SD_CMD | G9 | 32 | D, IN/OUT | SDIO command line |
| USB Interface | | | | |
| DP | A9 | 67 | D, IN/OUT | USB D+ |
| DM | A7 | 69 | D, IN/OUT | USB D- |
| TXRTUNE | B8 | 65 | A, IN | Tuning resistor input |
| Reset | | | | |
| RESETN | G7 | 23 | D, IN | Active-low chip reset input |
| GPIO | | | | |
| GPIO[0:31] | - | - | D, IN/OUT | Fully configurable GPIO pins (see pin multiplexing section) |
| Crystal | | | | |
| XTAL_I/MASTER_CLK | A11 | 63 | D, IN | Crystal In pin Host CLK In |
| XTAL_O | A13 | 61 | D, I/O | Crystal Out pin |
| Digital Microphones | | | | |
| DM0_CLK/CLKOUT | F12 | 44 | D, OUT | DMIC0 clock |
| DM0_DATA | H12 | 42 | D, IN | DMIC0 data |
| DM1_CLK | G11 | 29 | D, OUT | DMIC1 clock |
| DM1_DATA | J11 | 31 | D, IN | DMIC1 data |
| DM2_CLK | K12 | 33 | D, OUT | DMIC2 clock |
| DM2_DATA | K10 | 35 | D, IN | DMIC2 data |
| DM3_CLK | H10 | 25 | D, OUT | DMIC3 clock |

| Signal | Ball | Pad | Type, Direction | Description |
|---------------------------|------|-----|-----------------|--|
| DM3_DATA | J9 | 27 | D, IN | DMIC3 data |
| DM3_CLK_b | F14 | 43 | D, OUT | DMIC3 clock (alternate pin) |
| DM3_DATA_b | K16 | 39 | D, IN | DMIC3 data (alternate pin) |
| Analog Microphones | | | | |
| MIC0_P | F4 | 12 | A, IN | Mic0 input (positive) MADC input |
| MIC0_N | G3 | 14 | A, IN | Mic0 input (negative) |
| MIC1_P | J5 | 20 | A, IN | Mic1 input (positive) |
| MIC1_N | J3 | 18 | A, IN | Mic1 input (negative) |
| MIC2_P | J1 | 19 | A, IN | Mic2 input (positive) |
| MIC2_N | K2 | 21 | A, IN | Mic2 input (negative) |
| MIC3_P | K6 | 22 | A, IN | Mic3 input (positive) |
| MIC3_N | K4 | 24 | A, IN | Mic3 input (negative) |
| MIC_BIAS0 | G5 | 8 | A, OUT | Bias voltage for microphones |
| MIC_BIAS1 | F6 | 10 | A, OUT | Bias voltage for microphones |
| Multi-button Sense | | | | |
| BUTTON_SENSE | H6 | 6 | A, IN | Pushbutton array MADC input. |
| Headphone | | | | |
| HPLP | E1 | 13 | Diff, O | Headphone positive left output |
| HPLN | F2 | 11 | Diff, O | Headphone negative left output |
| HPRP | C1 | 5 | Diff, O | Headphone positive right output |
| HPRN | D2 | 7 | Diff, O | Headphone negative right output |
| VDDDRV_HP | E3 | 9 | A, OUT | Headphone driver supply filter pin |
| PWM LEDs | | | | |
| LED_SINK0 | B10 | 66 | D, OUT | PWM LED sink 0 1-Wire PLC TX (UART2_TX_b) |
| LED_SINK1 | C9 | 64 | D, OUT | PWM LED sink 1 |
| LED_SINK2 | D10 | 62 | D, OUT | PWM LED sink 2 |
| PWM3 | G9 | 32 | D, OUT | PWM LED output 3 |
| PWM4 | J7 | 28 | D, OUT | PWM LED output 4 |
| PWM5 | K8 | 30 | D, OUT | PWM LED output 5 |
| Power | | | | |
| PMU_EN | C7 | 72 | D, IN | Enable signal for PMU (LDOs, DC-DC) |
| POR_BYPASS | D8 | 2 | — | — |
| VSYS | A5 | 73 | P, IN | Main system supply input (from VBUS or battery). Input to DC-DC Input to 3.3-V LDO |
| SW_DCDC | A3 | 75 | P, OUT | Switching node of DC-DC converter |
| VDD_LDO | B2 | 74 | P, IN | Input to core LDO |

| Signal | Ball | Pad | Type, Direction | Description |
|----------|-----------------|----------------|-----------------|--|
| | | | | Input to 0.8-V LDO |
| VDD18 | C3 | 1 | P, IN | HP driver supply I/O voltage supply |
| VDD33 | B6 | 70 | P, OUT | Output of 3.3-V LDO |
| AVDD16 | E5 | 4 | P, OUT | Output of 1.65-V LDO |
| AVDD08 | C5 | 3 | P, OUT | Output of 0.8-V LDO |
| PVDDC | B4 | 71 | P, OUT | Output of core LDO |
| VDD | E11 F16 | 46 52 | P, IN | Digital cores voltage supply Power from PVDDC |
| VREF | G1 | 15 | A, OUT | 0.96 V analog reference voltage |
| VREF_GND | H2 | 17 | A, GND | Analog reference ground |
| AVSS | H4 | 16 | GND | Analog ground |
| VSS | D4 D6 G13 | 26 45 68 | GND | Digital ground |
| VSS_DCDC | A1 | 76 | GND | DC-DC ground |

3.2. Pin Assignments

The SR80 Series WLCSP pin assignment is shown in Figure 2.

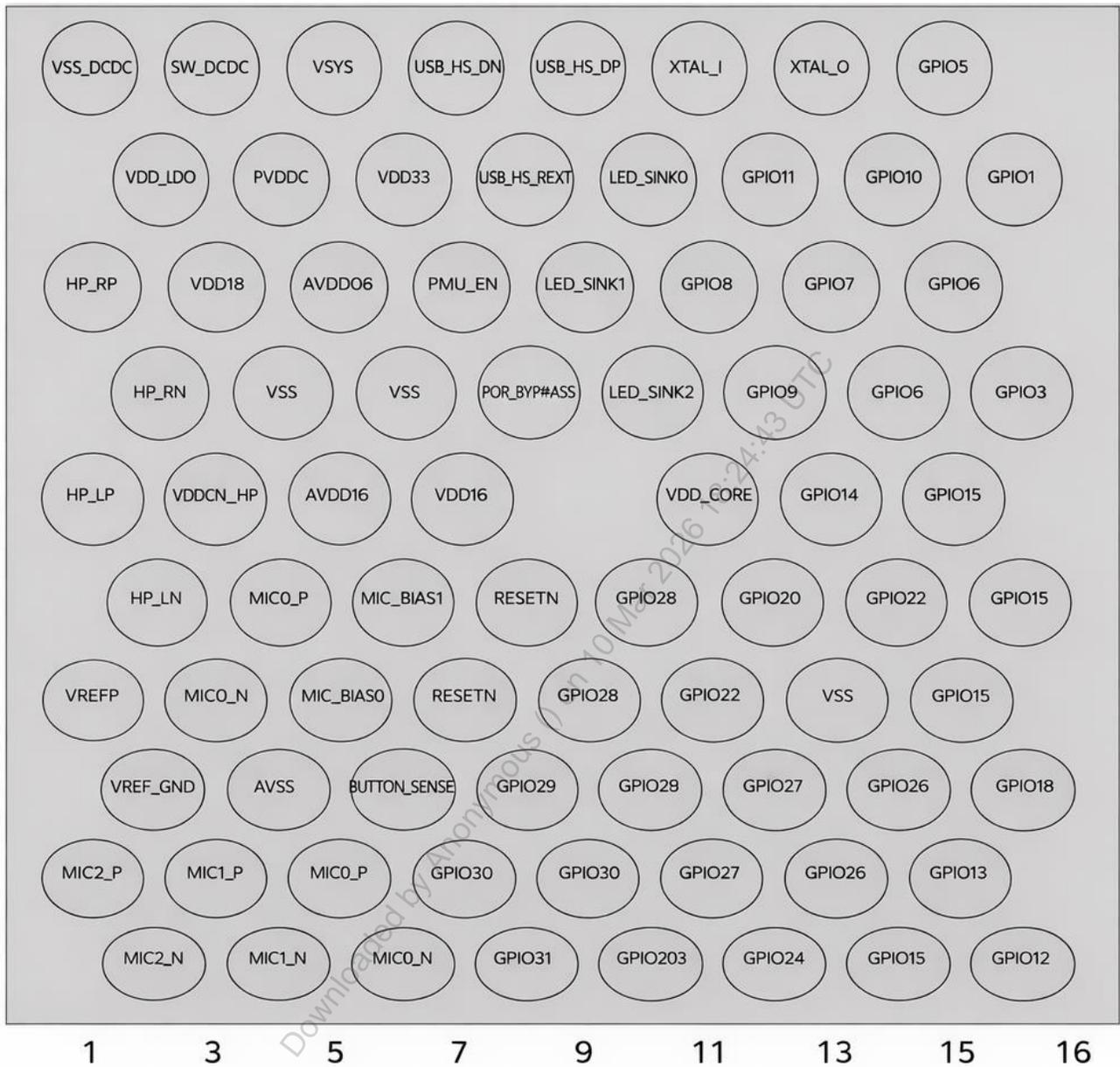


Figure 2. SR80 Series pin assignment (package balls facing down)

The SR80 Series LGA pin assignment is shown in Figure 3.

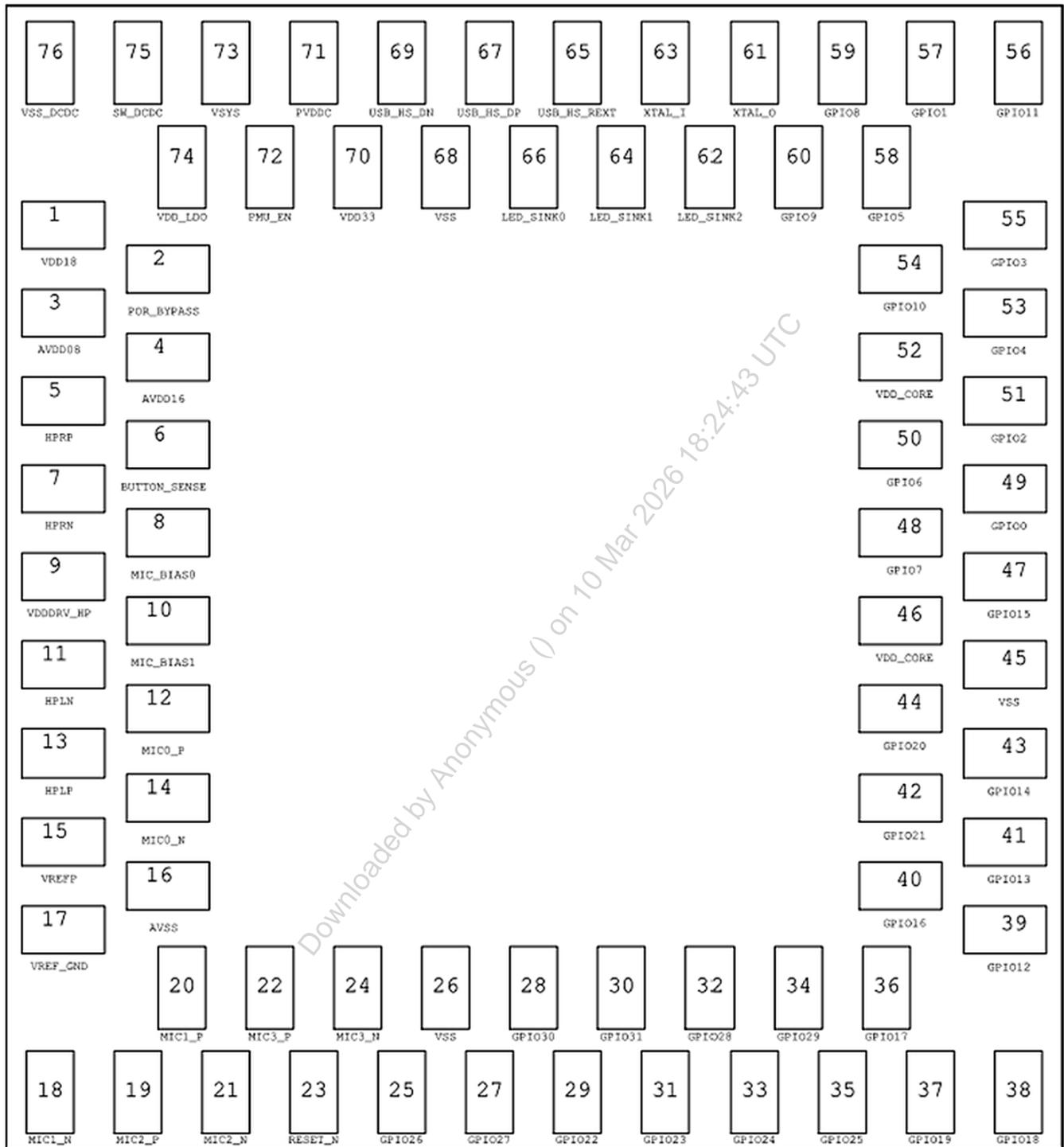


Figure 3. SR80 Series pin assignment (package pads facing down)

3.3. Pin Multiplexing

Many of the SR80 Series device pins support multiple functions, one of which can be enabled at a given time.

All 32 GPIOs can be configured for multiple functions (that is, functions are multiplexed at the pin). This section tabulates the functions supported by each of the device pins, along with other pin attributes (default function, input, output, pull-up/pull-down resistor and so on).

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Table 4. Pin multiplexing for the SR80 Series GPIOs.

| Name | Ball No. | Pad No. | MUX0 | MUX1 | MUX2 | MUX3 | MUX4 | MUX5 | MUX6 | PU/PD |
|----------------------------------|----------|---------|--------|--------------|---------------|---------------|---------------|------|------|----------------|
| QSPI_CS/SPI_SL_CS | E13 | 49 | GPIO0 | QSPI_CS | SPI_SL_CS | – | – | – | – | PU |
| QSPI_CLK/SPI_SL_SCK | B16 | 57 | GPIO1 | QSPI_CLK | SPI_SL_SCK | – | – | – | – | PU |
| QSPI_SI/QSPI_SIO0 /SPI_SL_SDI | D16 | 55 | GPIO3 | QSPI_SI/SIO0 | SPI_SL_MOSI | – | – | – | – | Hi-Z |
| QSPI_SO/QSPI_SIO1 /SPI_SL_SDO | C15 | 51 | GPIO2 | QSPI_SO/SIO1 | SPI_SL_MISO | – | – | – | – | Hi-Z |
| QSPI_SIO2/GP4 /UART1_TX | D14 | 53 | GPIO4 | QSPI_SIO2 | UART1_TX | – | – | – | – | Hi-Z |
| QSPI_SIO3/GP5 /UART1_RX | A15 | 58 | GPIO5 | QSPI_SIO3 | UART1_RX | – | – | – | – | Hi-Z |
| I2C_SLV_SCL/UART0_RX | D12 | 50 | GPIO6 | I2C_SLV_SCL | UART0_RX | I3C_SCL | SPI_SL_SCK_B | – | – | PU |
| I2C_SLV_SDA/UART0_TX | C13 | 48 | GPIO7 | I2C_SLV_SDA | UART0_TX | I3C_SDA | SPI_SL_MOSI_B | – | – | PU |
| I2SO_PCM_BCLK | E15 | 59 | GPIO8 | I2SO_BCLK | – | – | – | – | – | Hi-Z |
| I2SO_PCM_WS | C11 | 60 | GPIO9 | I2SO_WS | – | – | – | – | – | Hi-Z |
| I2SO_PCM_TX_DAT | B14 | 54 | GPIO10 | I2SO_TX_DAT | – | – | – | – | – | Hi-Z |
| I2SO_PCM_RX_DAT | B12 | 56 | GPIO11 | I2SO_RX_DAT | – | – | – | – | – | Hi-Z |
| GP12/TDI_A/I2S1_RX | K16 | 39 | TDI_A | GPIO12 | I2S1_RX_DAT | SPI_SL_CS_B | DM3_DATA_B | – | – | Hi-Z |
| GP13/I2S1_WS | K14 | 41 | GPIO13 | I2S1_WS | – | – | – | – | – | Hi-Z |
| GP14/TDO_A/I2S1_TX_DAT | F14 | 43 | TDO_A | GPIO14 | I2S1_TX_DAT | SPI_SL_MISO_B | DM3_CLK_B | – | – | PD (STRAPO) |
| GP15/I2S1_BCLK/TRSTN | G15 | 47 | GPIO15 | I2S1_BCLK | TRSTN | – | – | – | – | Hi-Z |
| TCK/I2C1_MS_SCL /UART2_TX | H16 | 40 | TCK_A | GPIO16 | I2C1_MS_SCL | UART2_TX | – | – | – | PU |
| TMS/I2C1_MS_SDA /UART2_RX | H14 | 36 | TMS_A | GPIO17 | I2C1_MS_SDA | UART2_RX | – | – | – | PU |
| I2CO_MS_SCL/I2S2_TX_B | J15 | 38 | GPIO18 | I2CO_MS_SCL | I2S1_TX_DAT_B | UART1_TX_B | I3C_SCL_B | – | – | PU |
| I2CO_MS_SDA/I2S2_RX_B | J13 | 37 | GPIO19 | I2CO_MS_SDA | I2S1_RX_DAT_B | UART1_RX_B | I3C_SDA_B | – | – | PU |

| Name | Ball No. | Pad No. | MUX0 | MUX1 | MUX2 | MUX3 | MUX4 | MUX5 | MUX6 | PU/PD |
|---|----------|---------|--------|---------------|------------|-------------|----------------|----------------|---------|----------------|
| DM0_CLK/CLKOUT | F12 | 44 | GPIO20 | DM0_CLK | CLKOUT | TDO_B | — | — | — | PU |
| DM0_DATA | H12 | 42 | GPIO21 | DM0_DATA | TDI_B | 1-Wire_RX_B | — | — | — | Hi-Z |
| DM1_CLK/UART0_TX_B /I2S2_CLK | G11 | 29 | GPIO22 | DM1_CLK | UART0_TX_B | I2S2_BCLK | SPI1_MSTR_CLK | — | — | Hi-Z |
| DM1_DATA/UART0_RX_B /I2S2_WS | J11 | 31 | GPIO23 | DM1_DATA | UART0_RX_B | I2S2_WS | SPI1_MSTR_CS | — | — | Hi-Z |
| DM2_CLK/CLKOUT UART0_CTS/I2S2_TX (STRAP1) | K12 | 33 | GPIO24 | DM2_CLK | UART0_CTS | I2S2_TX_DAT | SPI1_MSTR_MOSI | — | — | PD (STRAP1) |
| DM2_DATA/UART_RTS /TDI_B/I2S2_RX | K10 | 35 | GPIO25 | DM2_DATA | UART0_RTS | I2S2_RX_DAT | SPI1_MSTR_MISO | — | — | Hi-Z |
| GP26 (TEST_N) | H10 | 25 | GPIO26 | I2S2_BCLK_B | UART0_TX_C | DM3_CLK_A | — | SPI2_MSTR_CLK | SD_CLK | PU (TEST_N) |
| GP27 | J9 | 27 | GPIO27 | I2S2_WS_B | UART0_RX_C | DM3_DATA_A | — | SPI2_MSTR_CS | SD_DAT3 | Hi-Z |
| GP28 | G9 | 32 | GPIO28 | I2S2_TX_DAT_B | PWM3 | — | — | SPI2_MSTR_MOSI | SD_CMD | PD |
| GP29 | H8 | 34 | GPIO29 | I2S2_RX_DAT_B | — | — | — | SPI2_MSTR_MISO | SD_DAT0 | Hi-Z |
| GP30 | J7 | 28 | GPIO30 | I2C1_MS_SCL_B | PWM4 | — | — | UART0_CTS_B | SD_DAT1 | Hi-Z |
| GP31 | K8 | 30 | GPIO31 | I2C1_MS_SDA_B | PWM5 | — | — | UART0_RTS_B | SD_DAT2 | Hi-Z |

3.4. Bootstrapping Pins

Some signal pins in the SR80 Series device are used to configure the device for a certain mode at power up.

Table 5. Bootstrap pins.

| Bootstrap | Assigned Pin | Reset State | Description |
|-----------|--------------|-------------|--|
| STRAPO | GPIO14 | PD | STRAP[1:0] used by boot ROM to select boot option. |
| STRAP1 | GPIO24 | PD | |
| TEST_N | GPIO26 | PU | TEST_N is pulled down to enable product test mode; pull up at power up during normal mode. |

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4. Electrical Characteristics

4.1. Absolute Maximum Ratings

Absolute maximum ratings are those values beyond which permanent damage to the device may occur. These are stress ratings only, and functional operation under these conditions or at any other condition beyond those indicated for normal operation is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Table 6. Absolute maximum ratings.

| Supply | Min. | Max. | Units | Description |
|--------|------|------|-------|--|
| VSYS | -0.3 | 7.0 | V | Main system supply, typically from battery or USB interface. Input to 3.3-V LDO Input to DC-DC |
| VDD18 | -0.3 | 1.98 | V | VDD for HP driver, typically sourced from DC-DC I/O input supply, typically sourced from DC-DC |

4.2. DC Supply Voltages

Table 7. DC supply voltages

| Supply | Min. | Typical | Max. | Units | Description |
|-----------|------|------------|------|-------|--|
| VSYS | 3.1 | 3.7/5.0 | 5.5 | V | Main system supply, typically from battery or USB interface. Input to 3.3-V LDO Input to DC-DC |
| VDD33 | -5% | 3.3 | +5% | V | Output of 3.3-V LDO |
| VDD18 | -5% | 1.8 | +5% | V | VDD for HP driver, typically sourced from DC-DC I/O input supply, typically sourced from DC-DC |
| VDD_LDO | -5% | 1.8 1.0 | +5% | V | Input to core LDOs, typically source from DC-DC 1.8 V or 1.0 V depends on mode |
| PVDDC | -5% | 0.8 | +5% | V | Output of VDDCORE LDO |
| VDD | -5% | 0.8 | +5% | V | Input voltage supply for digital cores Power from PVDDC |
| AVDD16 | -5% | 1.6 | +5% | V | Output of 1.6-V LDO Main supply for analog TX/RX blocks |
| AVDD08 | -5% | 0.8 | +5% | V | Output of 0.8-V LDO Internally generated voltage for PLLs, OSC |
| VDDDRV_HP | -5% | 1.8 | +5% | V | Headphone driver supply filter pin |

4.3. Analog Characteristics

4.3.1. DAC Performance

Table 8. DAC analog characteristics

| DAC Output Path | | | | | |
|----------------------------------|---|---------|------------|---------|------------|
| Headphone load RL = 32 Ω | | | | | |
| Parameter | Test Conditions | Minimum | Typical | Maximum | Unit |
| Full scale output | RL = 16 Ω , THD < 0.1% RL = 32 Ω , THD < 0.1% | — | 0.7 1.0 | — | Vrms |
| Headphone load | — | 16 | 32 | — | Ω |
| Dynamic range | -60 dBFS signal source, A-weighted | — | 110 | — | dB |
| SNR | 0 dBFS for signal measurement Silence for noise measurement, A-weighted | — | 120 | — | dB |
| THD+N | -1 dB signal source | — | -95 | — | dB |
| Crosstalk | 32 Ω stereo load | — | -100 | -95 | dB |
| Intermodulation Distortion (IMD) | 5 mW output level -8 dB 60 Hz, -20 dB 7 kHz | — | -85 | — | dB |
| DC Offset | Voffset | — | <50 | 100 | μ V DC |
| Pop Level | — | — | -75 | — | dBV |
| Frequency response | 20Hz to 20 kHz | — | .04 | .05 | dB |
| Phase error | Between two channels 997 Hz, 0 dBFS signals | — | — | .01 | degree |
| PSRR | @217Hz – Square wave, 100 mVpp, 12.5% duty cycle | 80 | — | — | dB |
| | @1 kHz – Sine wave | 80 | — | — | dB |
| Output Capacitive Loading | Single-ended capacitance | — | 0.5 | — | nF |
| Output resistance | Measure -20 dBFS, 1 kHz signal with and without resistive load, calculate resistance. | — | 0.5 | — | Ω |

4.3.2. ADC Performance

Table 9. ADC analog characteristics

| MIC ADC Path Differential Cap-less Input | | | | | |
|---|--|---------|--------------------|---------|----------|
| Parameter | Test Conditions | Minimum | Typical | Maximum | Unit |
| Full scale voltage | 0 dB gain +6 dB gain +12 dB gain | — | 1.0 0.5 0.25 | — | V rms |
| Dynamic range | -60 dBFS, A weighted, 0 dB gain -60 dBFS, A weighted, 6 dB gain -60 dBFS, A weighted, 12 dB gain | — | 105 — — | — | dB |
| THD+N | -1 dBFS | — | -90 | — | dB |
| Crosstalk* | Channel to channel @ 1 kHz | — | -130 | -120 | dB |
| PSRR | 217 Hz, 200 mVpp | — | — | — | dB |
| Analog gain range | — | 0 | — | 12 | dB |
| Analog gain step size | — | — | 6 | — | dB |
| Input impedance | — | — | >1M | — | Ω |
| Input capacitance | — | — | 100 | — | pF |

* Crosstalk: Measurements are made with one channel driven to -1 dBFS and all other MIC inputs shorted to GND.

4.3.3. Microphone Bias Characteristics

Table 10. Microphone bias parameters

| Parameter | Test Conditions/Notes | Minimum | Typical | Maximum | Unit |
|-------------------------|--|---------|---------|---------|---------|
| Mic bias Output Voltage | Used as bias for the analog microphone | 1.2 | — | 3.0 | V |
| Mic bias Output Voltage | In bypass mode to 1.8 V supply | — | 1.8 | — | V |
| Mic bias Voltage Step | — | — | 0.1 | — | V |
| Mic bias Load Current | — | — | — | 5.0 | mA |
| Load Capacitance | Microphone bias output does not require load capacitor for stability | 0 | — | 10 | μ F |

4.4. Digital I/O

4.4.1. DC Characteristics

Table 11. DC characteristics – TTL compatible (GPIOs)

| Parameter | Symbol | Minimum | Typical | Maximum | Units | Notes |
|--|--------|--------------|---------|--------------|-------|-------|
| Input voltage low | VIL | 0 | – | 0.3 x DVDDIO | V | – |
| Input voltage high | VIH | DVDDIO x 0.7 | – | DVDDIO | V | – |
| Output voltage low | VOL | 0 | – | 0.2 x DVDDIO | V | – |
| Output voltage high | VOH | 0.8 x DVDDIO | – | DVDDIO | V | – |
| GPIO output sink current at VOL with DVDDIO = 1.8V | IOL | – | – | 8 | mA | – |
| GPIO output source current at VOH with DVDDIO = 1.8V | IOH | – | – | 8 | mA | – |

4.5. SD, SDIO Timing

4.5.1. SD, SDIO Default Mode Timing Parameters

Table 12. SD, SDIO Default Mode Timing Parameters

| Symbol | Parameter | Condition | Min | Typ | Max | Units |
|---|-------------------------------------|---------------------|-----|-----|-----|-------|
| f _{PP} | Clock Frequency Data Transfer Mode | – | 0 | 25 | 25 | MHz |
| f _{OD} | Clock Frequency Identification Mode | – | 0 | – | 400 | kHz |
| t _{WL} | Clock Low time | – | 10 | – | – | ns |
| t _{WH} | Clock High time | – | 10 | – | – | |
| t _{TLH} | Clock Rise time | – | – | – | 10 | |
| t _{THL} | Clock Fall time | – | – | – | 10 | |
| Inputs CMD, DAT (referenced to Clock): | | | | | | |
| t _{ISU} | Input Setup time | – | 5 | – | – | ns |
| t _{IH} | Input Hold time | – | 5 | – | – | |
| Outputs CMD, DAT (referenced to Clock): | | | | | | |
| t _{ODLY} | Output delay time | Data Transfer Mode | 0 | – | 14 | ns |
| t _{ODLY} | Output delay time | Identification Mode | 0 | – | 50 | |

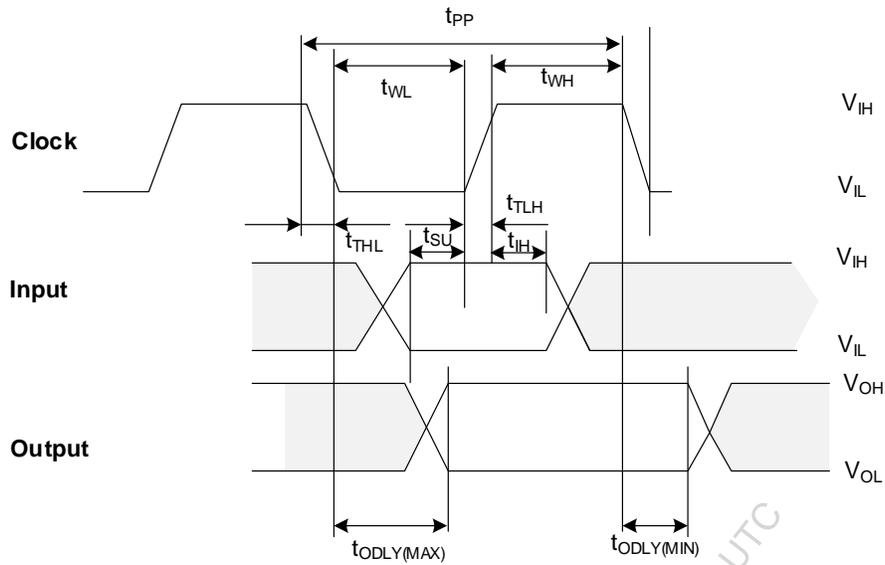


Figure 4. Timing Diagram Data Input/Output Referenced to Clock (Default)

4.5.2. SD, SDIO High-Speed Mode Timing Parameters

Table 13. SD, SDIO High-Speed Mode Timing Parameters

| Symbol | Parameter | Condition | Min | Typ | Max | Units |
|---|------------------------------------|--------------------|-----|-----|-----|-------|
| f_{PP} | Clock Frequency Data Transfer Mode | — | 0 | 50 | 50 | MHz |
| t_{WL} | Clock Low time | — | 7 | — | — | ns |
| t_{WH} | Clock High time | — | 7 | — | — | |
| t_{TLH} | Clock Rise time | — | — | — | 3 | |
| t_{THL} | Clock Fall time | — | — | — | 3 | |
| Inputs DAT (referenced to Clock): | | | | | | |
| t_{ISU} | Input Setup time | — | 6 | — | — | ns |
| t_{IH} | Input Hold time | — | 2 | — | — | |
| Outputs CMD, DAT (referenced to Clock): | | | | | | |
| t_{ODLY} | Output Delay time | Data Transfer mode | 0 | — | 14 | ns |
| t_{OH} | Output Hold time | — | 2.5 | — | — | |

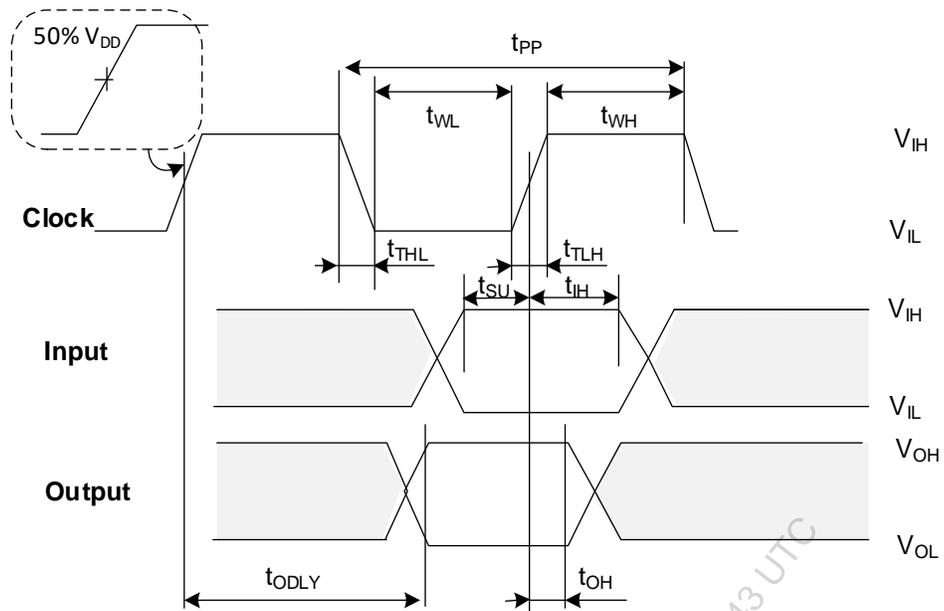


Figure 5. Timing Diagram Data Input/Output Referenced to Clock (high-speed)

4.5.3. SD, SDIO SDR50 Mode Timing Parameters

Table 14. SD, SDIO SDR50 Mode Timing Parameters

| Symbol | Parameter | | Condition | Min | Typ | Max | Units |
|---|------------------------------------|--------------------|-----------|------|-----|-----|-------|
| f_{PP} | Clock Frequency Data Transfer Mode | — | — | 0 | 100 | 100 | MHz |
| t_{WL} | Clock Low time | — | — | 1.44 | — | — | ns |
| t_{WH} | Clock High time | — | — | 1.44 | — | — | |
| t_{TLH} | Clock Rise time | — | — | — | — | 2 | |
| t_{THL} | Clock Fall time | — | — | — | — | 2 | |
| Inputs DAT (referenced to Clock): | | | | | | | |
| t_{ISU} | Input Setup time | — | — | 3 | — | — | ns |
| t_{IH} | Input Hold time | — | — | 0.8 | — | — | |
| Outputs CMD, DAT (referenced to Clock): | | | | | | | |
| t_{ODLY} | Output Delay time | Data Transfer mode | | 0 | — | 7.5 | ns |
| t_{OH} | Output Hold time | — | | 1.5 | — | — | |

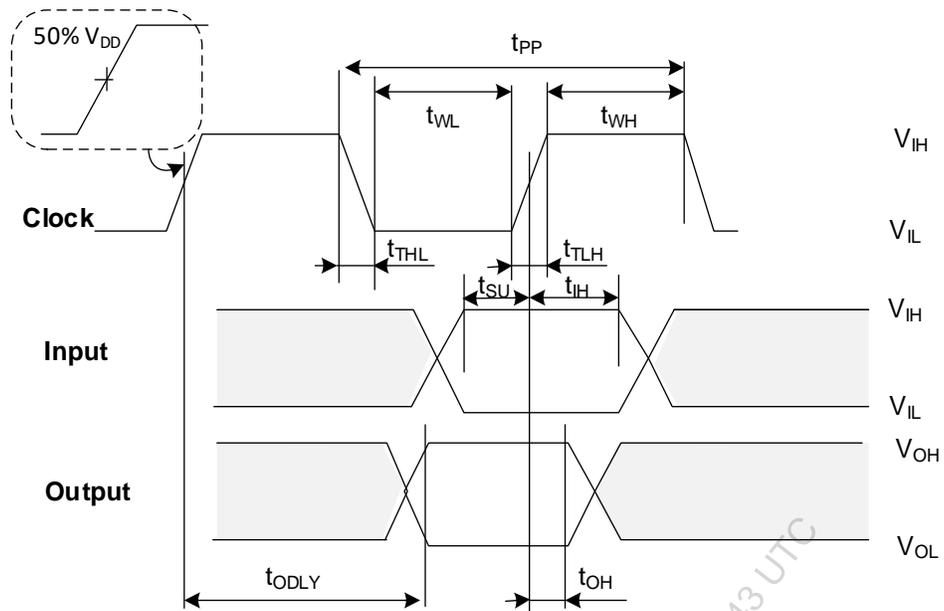


Figure 6. Timing diagram data I/O referenced to clock (high-speed & SDR50 mode)

4.6. Two-Wire Serial Interface (TWSI) Timing – I2C Compatible Timing

4.6.1. TWSI Standard and Fast Mode Timing

Table 15. TWSI Standard and Fast Mode Timing

| Symbol | Parameter | Condition | Min | Typ | Max | Units |
|--------------------------|---|-----------|------|-----|------|-------|
| F _{TWSI_SCL} | SCL Clock Frequency | 100 kHz | — | — | 100 | kHz |
| | | 400 kHz | — | — | 400 | |
| T _{TWSI_NS} | Noise Suppression Time at SCL, SDA Inputs | 100 kHz | — | — | 80 | ns |
| | | 400 kHz | — | — | 80 | |
| T _{TWSI_R} | SCL, SDA Rise time | 100 kHz | — | — | 1000 | |
| | | 400 kHz | — | — | 300 | |
| T _{TWSI_F} | SCL, SDA Fall Time | 100 kHz | — | — | 300 | |
| | | 400 kHz | — | — | 300 | |
| T _{TWSI_HIGH} | Clock High Period | 100 kHz | 4000 | — | — | |
| | | 400 kHz | 600 | — | — | |
| T _{TWSI_LOW} | Clock Low Period | 100 kHz | 4700 | — | — | |
| | | 400 kHz | 1300 | — | — | |
| T _{TWSI_SU:STA} | Start Condition Setup Time (for a Repeated Start Condition) | 100 kHz | 4700 | — | — | |
| | | 400 kHz | 600 | — | — | |
| T _{TWSI_HD:STA} | Start Condition Hold Time | 100 kHz | 4000 | — | — | |
| | | 400 kHz | 600 | — | — | |
| T _{TWSI_SU:STO} | Stop Condition Setup Time | 100 kHz | 4000 | — | — | |
| | | 400 kHz | 600 | — | — | |

| Symbol | Parameter | Condition | Min | Typ | Max | Units |
|--------------------------|-------------------------------|-----------|------|-----|-----|-------|
| T _{TWSL_SU:DAT} | Data in Setup Time | 100 kHz | 250 | — | — | |
| | | 400 kHz | 100 | — | — | |
| T _{TWSL_HD:DAT} | Data in Hold Time | 100 kHz | 300 | — | — | |
| | | 400 kHz | 300 | — | — | |
| T _{TWSL_BUF} | Bus Free Time | 100 kHz | 4700 | — | — | |
| | | 400 kHz | 1300 | — | — | |
| T _{TWSL_DLY} | SCL Low to SDA Data Out Valid | 100 kHz | 40 | — | 200 | |
| | | 400 kHz | 40 | — | 200 | |

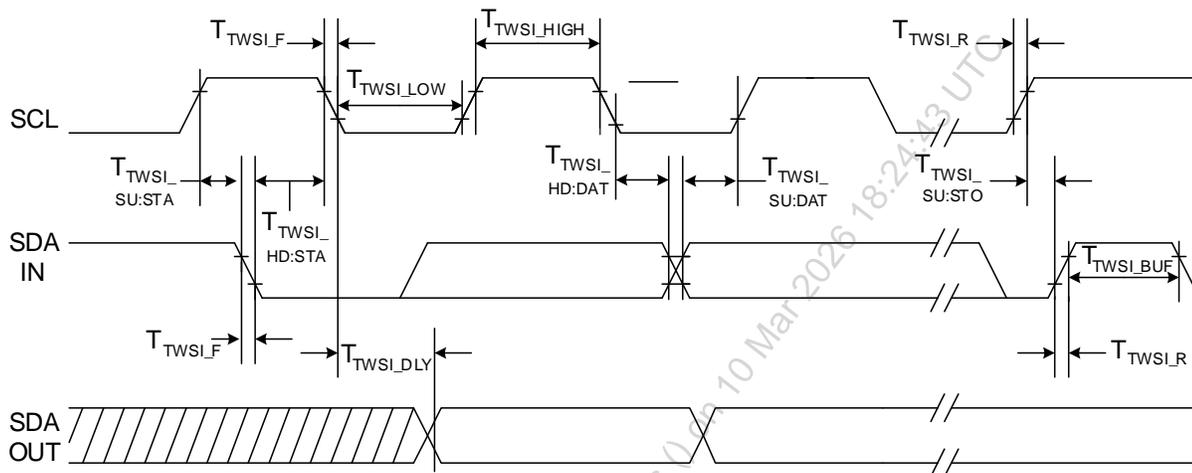


Figure 7. Two-wire serial interface timing

4.6.2. TWSI Fast mode plus – transfer rates up to 1 Mbits/s

For timing information, refer to the I2C Bus Specification.

4.7. UART Timing

Table 16. UART timing

| Symbol | Parameter | Condition | Min | Typ ¹ | Max | Units |
|--------|--------------|-----------|-----|------------------|-----|-------|
| — | Tx bit width | ±5% | — | 8.68 | — | μs |
| — | Rx bit width | ±5% | — | 8.68 | — | |

1. Typical values are given for a baud rate of 115.2 kbaud. Other baud rates are supported, and bit width scales inversely with the configured baud rate.

4.8. USB 2.0 Timing

The SR80 Series features an integrated USB 2.0 high-speed device interface (up to 480 Mbps). The interface supports USB Audio Class 1.0 and USB Audio Class 2.0 and provides up to 10 endpoints

4.8.1. USB 2.0 DC Characteristics

Table 17. USB 2.0 DC Electrical

| Symbol | Parameter | Condition | Min | Typ | Max | Units |
|---|--|---|-------|-----|------|-------|
| V _{IH} | High (driven) | Note1 | 2.0 | — | — | V |
| V _{IHZ} | High (floating) | | 2.7 | — | 3.6 | |
| V _{IL} | Low | | — | — | 0.8 | |
| V _{DI} | Differential Input Sensitivity | $(D^+) - (D^-)$ Note1 | 0.2 | — | — | V |
| V _{CM} | Differential Common Mode Range | Includes V _{DI} range Note1 | 0.8 | — | 2.5 | |
| Input Levels for High-speed | | | | | | |
| V _{HSSQ} | High-speed squelch detection threshold (differential signal amplitude) | — | 100 | — | 150 | mV |
| V _{HSDSC} | High-speed disconnect detection threshold (differential signal amplitude) | — | 525 | — | 625 | |
| V _{HSCM} | High-speed data signaling common mode voltage range (guideline for receiver) | — | -50 | — | 500 | |
| Output Levels for Full-speed | | | | | | |
| V _{OL} | Low | Note1, Note2 | 0.0 | — | 0.3 | V |
| V _{OH} | High (Driven) | Note1, Note3 | 2.8 | — | 3.6 | |
| V _{OSE1} | SE1 | — | 0.8 | — | — | |
| V _{CRS} | Output Signal Crossover voltage | Note4 | 1.3 | — | 2.0 | |
| Output Levels for High-speed | | | | | | |
| V _{HSOI} | High-speed idle level | — | -10.0 | — | 10.0 | mV |
| V _{HSOH} | High-speed data signaling high | — | 360 | — | 440 | |
| V _{HSOL} | High-speed data signaling low | — | -10.0 | — | 10.0 | |
| V _{CHIRPJ} | Chirp J level (differential voltage) | — | 700 | — | 1100 | |
| V _{CHIRPK} | Chirp K level (differential voltage) | — | -900 | — | -500 | |
| Input Capacitance for Full-speed | | | | | | |
| C _{IND} | Downstream Facing Port | Note5 | — | — | 150 | pF |
| C _{INUB} | Upstream Facing Port (without cable) | Note6 | — | — | 100 | |
| C _{EDGE} | Transceiver edge rate control capacitance | — | — | — | 75 | |

| Symbol | Parameter | Condition | Min | Typ | Max | Units |
|----------------------------------|--|--------------|-------|-----|-------|-------|
| Terminations | | | | | | |
| R _{PU} | Bus pull-up Resistor on Upstream facing port | 1.5 kOhm ±5% | 1.425 | — | 1.575 | kOhm |
| R _{PD} | Bus pull-down Resistor on Downstream Facing Port | 15 kOhm ±5% | 14.25 | — | 15.75 | |
| Z _{INP} | Input impedance exclusive of pullup/pull-down (for full-speed) | — | 300 | — | — | |
| V _{TERM} | Termination voltage for upstream facing port pull-up (RPU) | — | 3.0 | — | 3.6 | V |
| Termination in High-speed | | | | | | |
| V _{HSTERM} | Termination voltage in high-speed | — | -10 | — | 10 | mV |

1. Measured at A or B connector.
2. Measured with RL of 1.425 kohm to 3.6V.3.
3. Measured with RL of 14.25 kohm to GND.
4. Excluding the first transition from the idle state.
5. Measured at A receptacle.6.
6. Measured at B receptacle.

4.8.2. USB 2.0 Source Electrical Characteristics

Table 18. USB High-Speed Source Electrical Characteristics

| Symbol | Parameter | Condition | Min | Typ | Max | Units |
|-------------------------------|--|-----------|----------|-----|-----------------------|-------|
| Driver Characteristics | | | | | | |
| T _{HSR} | Rise Time (10%–90%) | — | 500 | — | — | ps |
| T _{HSF} | Fall Time (10%–90%) | — | 500 | — | — | |
| Z _{HSDRV} | Driver Output Resistance (which also serves as high-speed termination) | — | 40.5 | — | 49.5 | Ohm |
| Clock Timings | | | | | | |
| T _{HSDRAT} | High-speed Data Rate | — | 479.760 | — | 480.240 | Mbps |
| T _{HSFRAM} | Microframe Interval | — | 124.9375 | — | 125.0625 | µs |
| T _{HSRFI} | Consecutive Microframe Interval Difference | — | — | — | 4 highspeed bit times | — |

Table 19. USB Full-speed Source Electrical Characteristics

| Symbol | Parameter | Condition | Min | Typ | Max | Units |
|--------------------------------|---|---|---------|-----|---------|-------|
| Driver Characteristics | | | | | | |
| T _{FR} | Rise Time | — | 4 | — | 20 | ns |
| T _{FF} | Fall Time | — | 4 | — | 20 | |
| T _{FRFM} | Differential Rise and Fall Time Matching | T _{FR} /T _{FF} Note ¹ | 90 | — | 111.11 | % |
| Z _{DRV} | Driver Output Resistance for driver which is not high-speed capable. | — | 28 | — | 44 | Ohm |
| Clock Timings | | | | | | |
| T _{FDRATHS} | Full-speed Data Rate for hubs and devices which are high speed capable. | Average bit rate | 11.9940 | — | 12.0060 | Mbps |
| T _{FDRATE} | Full-speed Data Rate for devices which are high-speed capable. | Average bit rate | 11.9700 | — | 12.0300 | |
| T _{FRAME} | Frame Interval | — | 0.9995 | — | 1.0005 | ms |
| T _{RFI} | Consecutive Frame Interval Jitter | No clock adjustment | — | — | 42 | ns |
| Full-speed Data Timings | | | | | | |
| T _{DJ1} | Source Jitter Total (including frequency tolerance): To Next Transition | Note ¹ Note ² Note ³ | -3.5 | — | 3.5 | ns |
| T _{DJ2} | For Paired transitions | Note ⁴ | -4 | — | 4 | |
| T _{FDEOP} | Source Jitter for Differential Transition to SEO Transition | Note ³ | -2 | — | 5 | |
| T _{JR1} | Receiver jitter: To Next Transition | Note ³ | -18.5 | — | 18.5 | |
| T _{JR2} | For Paired Transitions | — | -9 | — | 9 | |
| T _{FEOPT} | Source SEO interval of EOP | — | 160 | — | 175 | |
| T _{FEOPR} | Receiver SEO interval of EOP | Note ⁵ | 82 | — | — | |
| T _{FST} | Width of SEO interval during differential transition | — | — | — | 14 | |

1. Excluding the first transition from the idle state.
2. Timing difference between the differential data signals.
3. Measured at crossover point of differential data signals.
4. For both transitions of differential signaling.
5. Must accept as valid EOP.

4.9. I2S Timing

4.9.1. I2S Host Mode Timing

Table 20. I2S Host Mode Timing Parameters

| Symbol | Parameter | Condition | Min | Typ | Max | Units |
|-------------|---|-----------|-----|------|--------|-------|
| F_{BCLK} | BCLK Frequency | — | — | — | 24.576 | MHz |
| F_s | Sample rate | — | — | — | 192 | kHz |
| D_{BCLK} | BCLK duty cycle | — | — | 50 | — | % |
| T_{SDPD1} | BCLK rising edge to SDATA output valid | — | — | 0.5 | — | ns |
| T_{LRPD} | BCLK rising edge to LRCK valid | — | — | 0.5 | — | ns |
| T_{SDS} | Set-up time SDATA input with regard to BCLK rising edge | — | — | 6.58 | — | ns |
| T_{SDH} | Hold time SDATA Input with regard to BCLK rising edge | — | — | 0 | — | ns |
| F_{MCLK} | MCLK (not shown) output frequency | — | — | — | 12.288 | MHz |
| D_{MCLK} | MCLK output duty cycle | — | — | 50 | — | % |

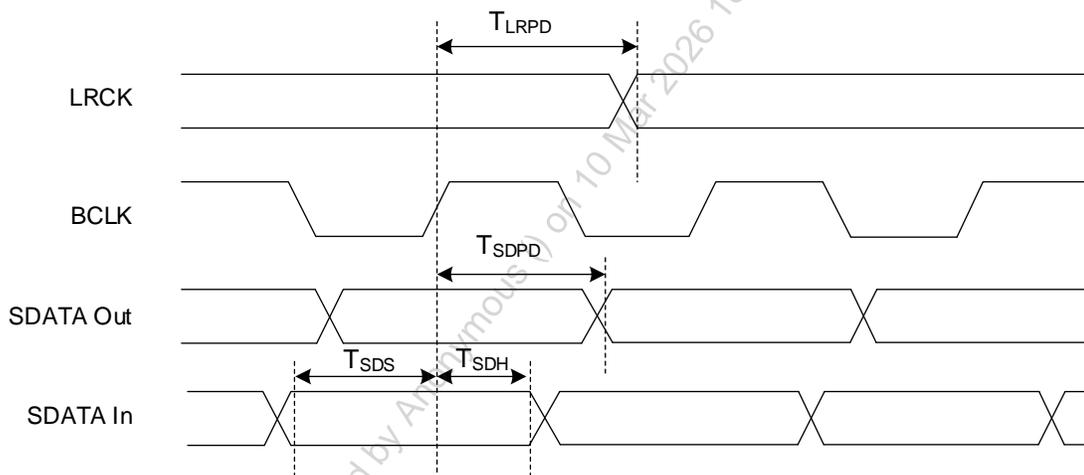


Figure 8. I2S Host mode timing.

4.9.2. I2S Target Mode Timing

Table 21. I2S Target mode timing parameters.

| Symbol | Parameter | Condition | Min | Typ | Max | Units |
|------------|--|-----------|-----|-----|--------|-------|
| F_{BCLK} | BCLK Frequency | — | — | — | 24.576 | MHz |
| F_s | Sample rate | — | — | — | 192 | kHz |
| D_{BCLK} | BCLK duty cycle | — | — | 50 | — | % |
| T_{LRS} | Setup time LRCK input with regard to BCLK active edge | — | — | 0.5 | — | ns |
| T_{LRH} | Hold time LRCK input with regard to BCLK active edge | — | — | 0.5 | — | ns |
| T_{SDS} | Setup time SDATA Input with regard to BCLK active edge | — | — | 2 | — | ns |
| T_{SDH} | Hold time SDATA Input with regard to BCLK active edge | — | — | 0 | — | ns |
| F_{MCLK} | MCLK (not shown) input frequency | — | — | — | 25 | MHz |
| D_{MCLK} | MCLK input duty cycle | — | — | 50 | — | % |

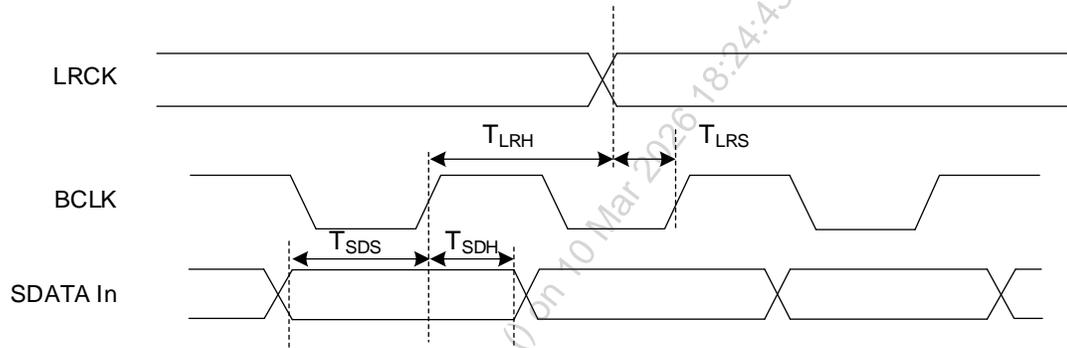


Figure 9. I2S Target mode timing.

4.10. SPI Timing

Table 22. SCLK Cycle Time Configurable Range

| Symbol | Parameter | Condition | Min | Typ | Max | Units |
|-------------------|-------------------------|--|-----|-----|---------|-------|
| T _{SCLK} | SoC SPI SCLK cycle time | 100 MHz SoC SPI controller input clock | 20 | — | 655,340 | ns |

(Applies across the full range of values listed in [Recommended Operating Conditions](#), unless otherwise specified.)

Table 23. Motorola SPI Mode 0/2 Timing

| Symbol | Parameter | Condition | Min | Typ | Max | Units |
|-------------------|---|--|------|-----|------|-------|
| T _{LS1} | Time from SSn assertion to the first SCLK active edge | The first SPI cycle in a transfer | — | 1.5 | — | TSCLK |
| | | Subsequent SPI cycles | — | 0.5 | — | |
| T _{LS2} | Time from the last SCLK inactive edge to SSn de-assertion | Other than the last SPI cycle | — | 1 | — | TSCLK |
| | | The last SPI cycle in a transfer | — | 1.0 | — | |
| T _{CH} | SCLK high time | — | — | 0.5 | — | TSCLK |
| T _{CL} | SCLK low time | — | — | 0.5 | — | TSCLK |
| T _{LH} | SSn de-assertion Time between SPI cycles | If Tx FIFO is not empty at the end of the previous SPI cycle | — | 0.5 | — | TSCLK |
| | | If Tx FIFO is empty | 2 | — | — | |
| T _{SET} | Setup time MISO with regard to SCLK active edge | — | 3.5 | — | — | ns |
| T _{HOLD} | Hold time MISO with regard to SCLK active edge | — | 0 | — | — | ns |
| T _{VAL1} | Time from SSn assertion to MOSI MSB valid | The first SPI cycle in a transfer | — | 1 | — | TSCLK |
| | | Subsequent SPI cycles | — | 0 | — | |
| T _{VAL2} | Time from SCLK inactive edge to MOSI data valid | — | 0.12 | — | 1.28 | ns |

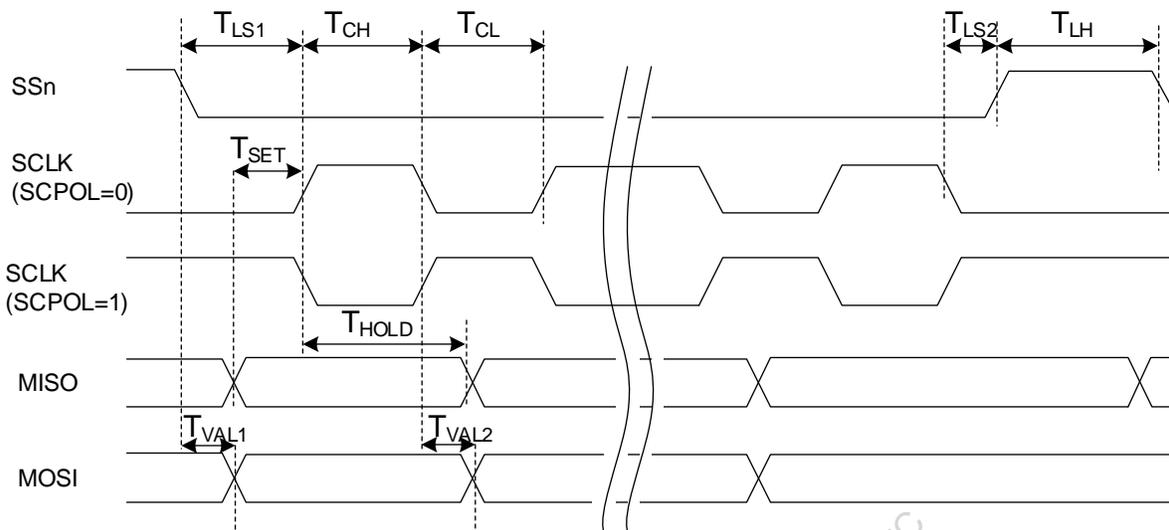


Figure 10. Motorola SPI Mode 0/2 (SCPH = 0)

(Applies across the full range of values listed in Recommended Operating Conditions, unless otherwise specified.)

Table 24. Motorola SPI Mode 1/3 Timing

| Symbol | Parameter | Condition | Min | Typ | Max | Units |
|------------|---|--|-----|-----|-----|-------|
| T_{LS1} | Time from SSn assertion to the first SCLK active edge | — | — | 1.0 | — | TSCLK |
| T_{LS2} | Time from the last SCLK inactive edge to SSn de-assertion | — | — | 1.0 | — | TSCLK |
| T_{CH} | SCLK high time | — | — | 0.5 | — | TSCLK |
| T_{CL} | SCLK low time | — | — | 0.5 | — | TSCLK |
| T_{LH} | SSn de-assertion Time between SPI cycles | If Tx FIFO is not empty at the end of the previous SPI cycle | — | 0 | — | TSCLK |
| | | If Tx FIFO is empty | 1.5 | — | — | |
| T_{SET} | Setup time MISO with regard to SCLK active edge | — | 3.5 | 30 | — | ns |
| T_{HOLD} | Hold time MISO with regard to SCLK active edge | — | — | 30 | — | ns |
| T_{VAL1} | Time from SSn assertion to MOSI MSB valid | The first SPI cycle in a transfer | — | 1 | — | TSCLK |
| | | Subsequent SPI cycles | — | 0 | — | |
| T_{VAL2} | Time from SCLK inactive edge to MOSI data valid | — | — | 0.5 | — | ns |

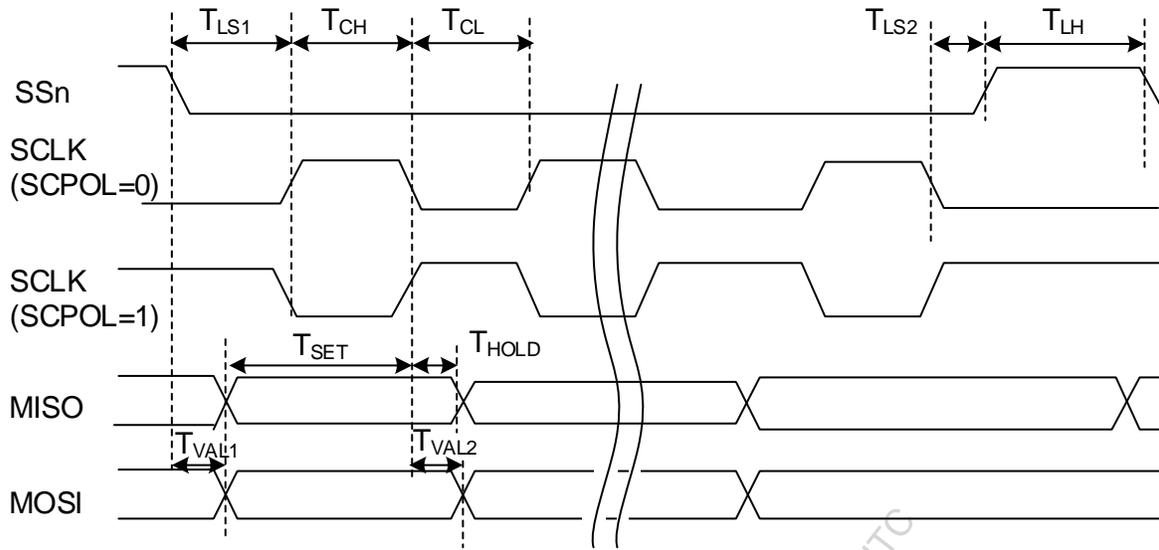


Figure 11. Motorola SPI Mode 1/3 (SCPH = 1)

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5. Codec Subsystem

5.1. Overview

A low-power, high-performance, audio codec subsystem provides all the needed circuitry to connect to external transducers.

The DACs and ADCs within the codec section convert the audio to digital streams that can be processed by one or more of the DSPs in the SR80 Series. Audio is processed in two primary paths, the ANC path, and the Audio/Voice processing path. The ANC path uses specialized filters, data paths, and processing for extremely low latency while the Audio/Voice Processing Path offers flexible use of the available resources. The CODEC can be split into two functional sections, the record path (often referred to as the *transmit* or *send path* in telephony) and the playback path (often referred to as the *receive path* in telephony).

The signal data paths are shown in Figure 12.

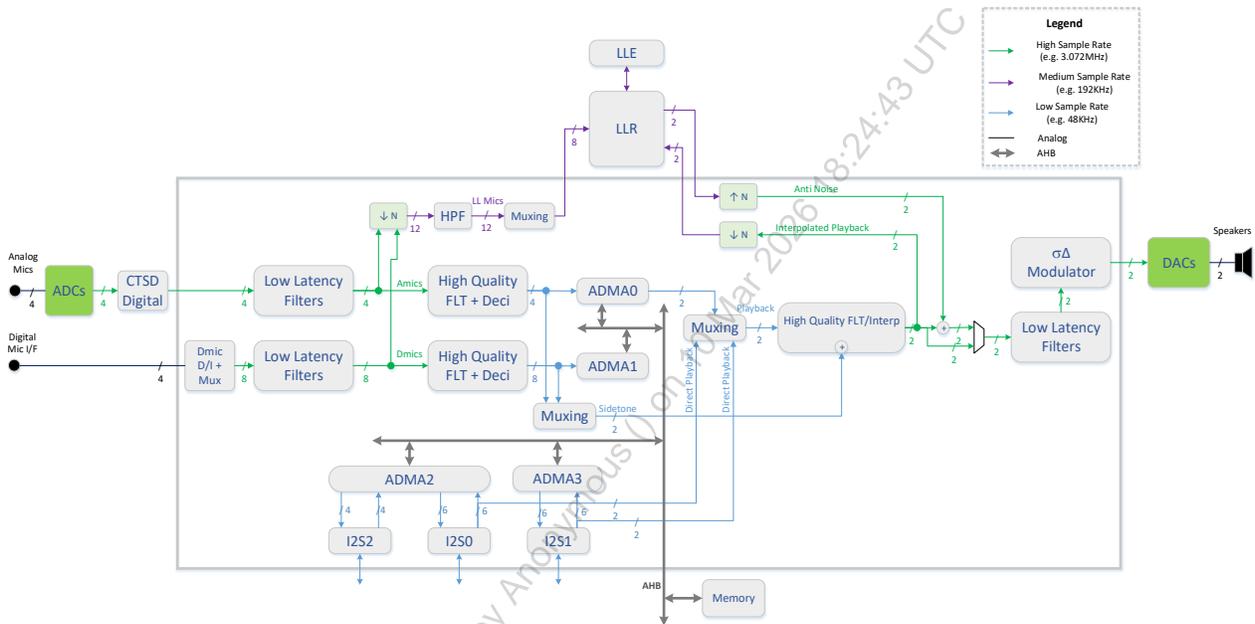


Figure 12. SR80 Series Audio data paths

5.2. Input Path

The codec subsystem has four analog inputs comprised of high dynamic range, high performance ADCs.

The codec subsystem also provides four PDM stereo DMIC interfaces, to connect to eight digital microphones.

5.2.1. Analog Inputs

The SR80 Series has 4 analog inputs capable of supporting microphone or line-level inputs.

- Each of the analog inputs has a dedicated high dynamic range continuous time sigma-delta (CTSD) ADC and programmable gain amplifier (PGA).
- Inputs can be used in a cap-less configuration, without the need for AC-coupling caps for BOM/area savings.
- Inputs provide analog gains of 0, 6, or 12 dB and support input levels up to 1.0 V.
- The analog input channels have dedicated input decimation filters, separate from the digital mic input channels.
- Low latency signal paths to the LLE engine enable ANC functionality.
- High-performance signals path enables advanced microphone DSP algorithms.

Two low noise microphone bias outputs provide 1.2 V to 3.0 V (100 mV resolution) and up to 5 mA of current to power attached microphones. The mic bias generation circuitry uses the 3.3 V LDO.

The mic bias voltage outputs can be used in a “no cap” configuration (0–1 nF) and “with cap” configuration (50 nF – 10 μ F), depending on system needs. Configuration setting is via control registers.

The mic bias outputs can also be used in bypass mode, where the 1.8 V supply is connected directly to the mic bias output. This can be used to increase power efficiency, since the 1.8 V supply is generated from a high efficiency DC-DC converter, whereas in normal (non-bypass) mode the mic bias voltage is generated from the VDD33 supply.

Input MICOP can be re-purposed as input to the 10-bit MADC, for general-purpose voltage measurements.

5.2.2. Microphone Bias

Two low noise microphone bias outputs provide 1.2 V to 3.0 V (100 mV resolution) and up to 5 mA of steady-state DC current to power attached microphones. The mic bias generation circuitry uses the 3.3 V LDO.

The Microphone Bias output current is limited to a value that is configurable by firmware (2 mA, 6 mA, 10 mA, or *disable*). Transient currents during 'output enable' can be higher than 5 mA depending on loading, so it's recommended to set a higher current limit, such as 10 mA, during this transition; the current limit can then be lowered to 6 mA or 2 mA.

The mic bias voltage outputs can be used in a “no cap” configuration (0–1 nF) and “with cap” configuration (50 nF – 10 μ F), depending on system needs. Configuration setting is via control registers.

The mic bias outputs can also be used in bypass mode, where the 1.8 V supply is connected directly to the mic bias output. This can be used to increase power efficiency, since the 1.8 V supply is generated from a high efficiency DC-DC converter, whereas in normal (non-bypass) mode the mic bias voltage is generated from the VDD33 supply.

Input MICOP can be re-purposed as input to the 10-bit MADC, for general-purpose voltage measurements.

5.2.3. Digital Microphone Inputs

The SR80 Series has four PDM digital microphone interfaces with separate clock and data pins. One or two digital microphones can be connected to each interface and a total of 8 digital microphones can be used at the same time.

One DMIC clock cannot be configured to be as the clock source for multiple DATA pins.

Each of the digital mic inputs has dedicated hardware decimation filters, to enable simultaneous use of all microphones (8 digital, 4 analog).

5.2.4. Input Path Signal Decimation

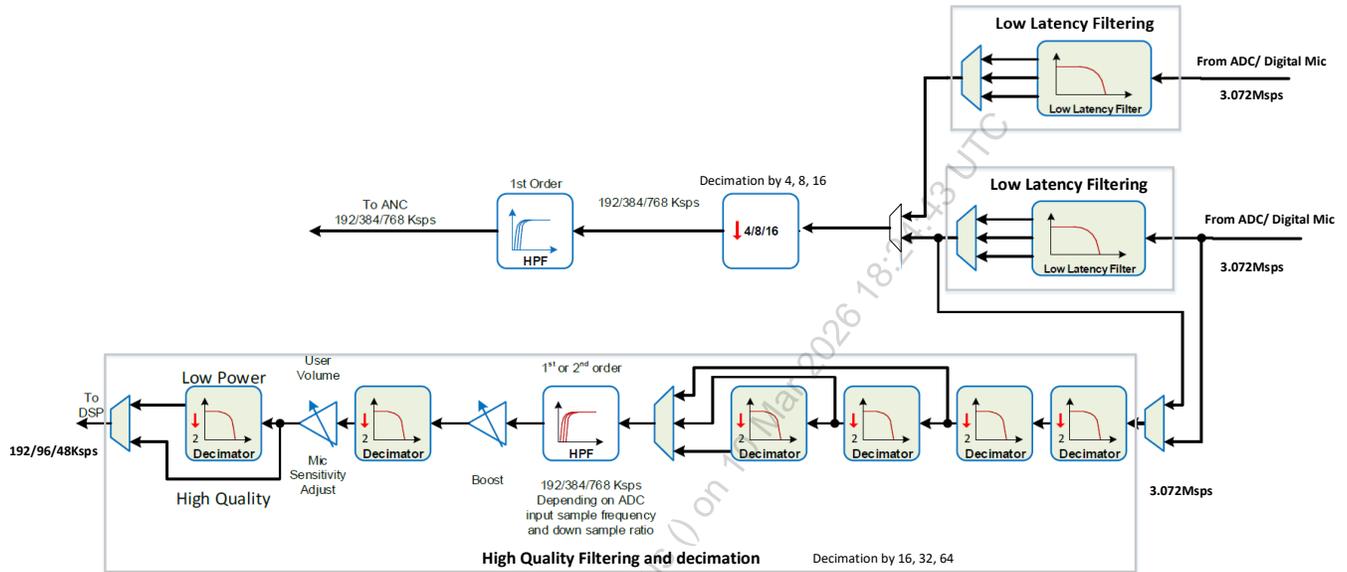


Figure 13. Input signal decimation in SR80 Series

5.3. Analog Playback Path

The Playback path provides hardware interpolation filters, a stereo high-performance DAC, and stereo class-AB/G headphone amplifier to connect the SR80 Series to headphones or other devices.

Low latency signal paths from/to the LLE engine enable ANC functionality.

High-performance audio paths support ANC as well as telephony and media applications. Volume control, anti-pop circuitry, and side-tone mixing are also provided.

5.3.1. DAC + Headphone Amplifier

The DAC and headphone amplifier are optimized for high-performance playback, using a class-AB design.

In addition, class G operation allows power-sensitive use cases to benefit from improved power efficiency by switching to a lower amplifier supply voltage.

The outputs are differential (balanced) and thus should not be connected to ground-connected loads.

- 1.0 V_{rms}/32 Ω
- 0.7 V_{rms}/16 Ω
- >40 kHz bandwidth (-1 dB)
- ± 0.03 dB magnitude response from 20 Hz to 20 kHz
- 24-bit DAC
- Dynamic range of 110 dBA
- Volume control: -74 to +6 dB in 0.5 dB steps
- Audio sample rates up to 192 kHz
- Low latency ANC support

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5.3.2. Playback Path Signal Interpolation

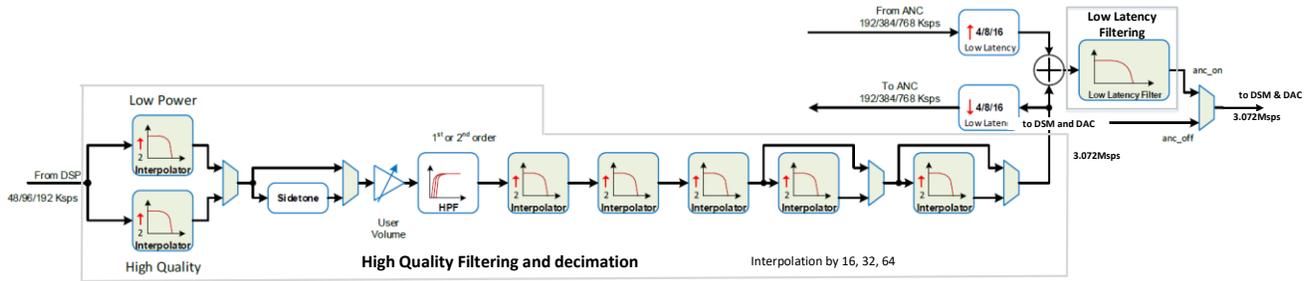


Figure 14. Playback path signal interpolation in SR80 Series

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6. M33 CPU Subsystem

A secure, single-core, 32-bit Arm Cortex-M33 based SSE200 sub-system provides system and sensor management and other functions.

The Cortex-M33 was developed to address all embedded and IoT markets, especially those requiring efficient security or digital signal control. TrustZone for ARMv8-M is the foundation of security for all embedded applications. The Cortex-M33 achieves an optimal blend between real time determinism, energy efficiency, software productivity, and system security that opens the door for many new applications and opportunities across diverse markets.

Supported functions:

- System management
- Trust management
- Sensor fusion
- Wake management
- Interrupt processing
- Firmware update
- I/O management

6.1. Hardware Features

The SSE200 system contains:

- A Cortex M33 32-bit processor running up to 250 MHz
- 64-bit wide AXI bus matrix
- Secured subsystem
- System controller
- Power, clock, and reset control infrastructure
- Interrupt controller
- Trust management and encryption

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6.2. Software Supported Features

6.2.1. Boot Host

The CPU subsystem boots while all other processors are held in reset allowing for a secure and orderly boot process. The remaining processors are initialized and managed by the CPU subsystem.

6.2.2. Sensor Hub

Connected sensors are managed by the M33 subsystem.

6.3. Memory

The CPU subsystem has:

- 16 KB I-Cache
- 16 KB D-Cache
- 128 KB local SRAM memory
- 64 KB I-tightly coupled memory
- 64 KB D-tightly coupled memory
- 96 KB Local ROM

The QSPI interface allows for code execution from external memory.

6.4. Test/Debug

6.4.1. UART Capture

Debug print messages can be output to a terminal, during code development.

6.4.2. Debug Expansion Module

The debug expansion module provides a trace port and a JTAG debug interface. The JTAG debug interface has these key features:

- JTAG port from DAPSWJDP
- Supports JTAG request to power up the PD_DEBUG power domain through the debug power integration kit (PIK)
- System timestamp generation

6.4.3. M33 Debug Interface

- Timestamp. A 64-bit input driven from an external timestamp generator. The subsystem routes the timestamp to the CPU elements, for use in the Cortex-M33 processor trace data.
- Debug Access. The DAP interface allows an external DAP to access the debug logic in the subsystem.
- Cross Trigger. The CTI interface provides some trigger inputs and outputs for system expansion.
- Debug expansion

6.5. Security

The following security components are available in SSE200 system.

- **MPC:** AHB5 TrustZone memory protection controller
- **AHB5 PPC:** AHB5 TrustZone peripheral protection controller
- **APB PPC:** APB TrustZone peripheral protection controller
- **MSC:** AHB5 TrustZone host security controller
- **SAU:** Security attribution unit
- **IDAU:** Implementation defined attribution unit

6.5.1. Memory Protection Controller

The memory protection controller (MPC) gates transactions to the AHB5 host interface when a security violation occurs. The controller can be instantiated in the system in connection to any non-security aware AHB5 memory. The security checking is done based on block/page level, configured externally by the security controller through an APB interface.

6.5.2. APB Peripheral Protection Controller Block

PPC gates transactions and responses from peripherals when a security violation occurs. The APB4 TrustZone peripheral protection controller can be instantiated in the system in connection to any non-security aware peripherals. The security checking is done based on peripheral (PSEL) inputs configured externally by configuration inputs. (Similar to AHB5 PPC).

6.5.3. MSC with Implementation Defined Attribution Unit

The MSC is an adapter used to connect a legacy AHB-lite host, or an AHB5 host without the security extension to an AHB5 system and add TrustZone for ARMv8-M capability. IDAU Lite interface is used to check if the address is secure or non-secure.

6.5.4. Security Attribution Unit

The MPU improves system reliability by defining the memory attributes for different memory Regions. It provides up to 16 different regions, and an optional predefined background region. There can be two MPUs, one secure and one non-secure. Each MPU can define memory attributes independently. The SAU improves system security by defining security attributes for different regions. It provides up to 8 different regions and a default background region.

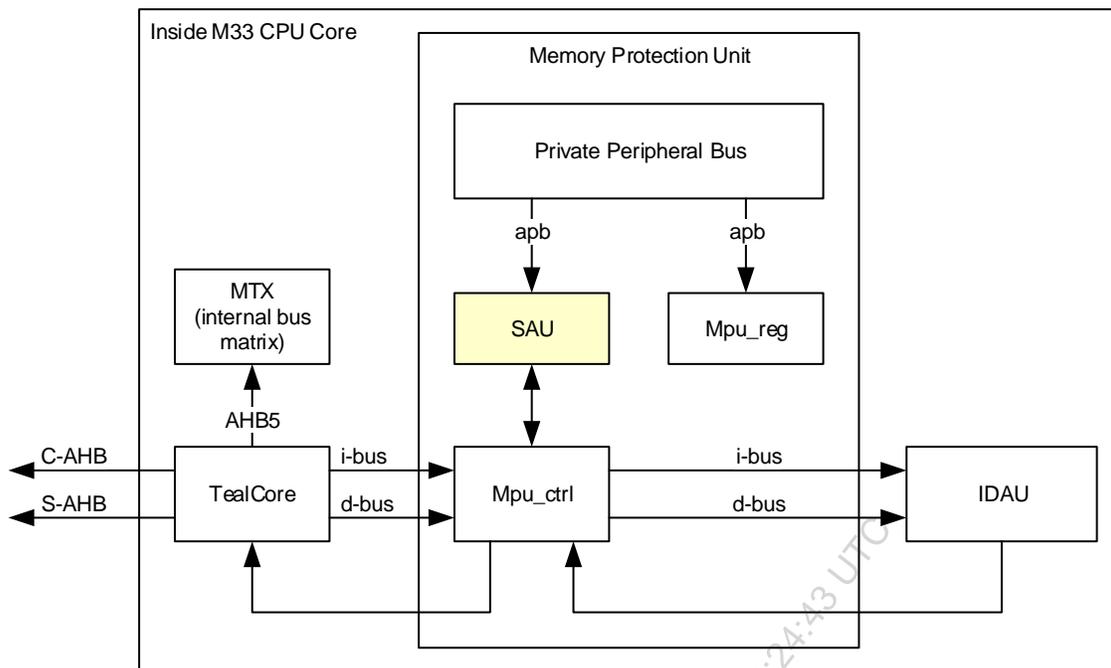


Figure 15. M33 CPU core

6.5.5. Security-use Cases

- Secure boot
 - OTP
 - AES
- Secure firmware upgrade
 - I²C in secure region (via Bluetooth)
- Device identification/authentication
 - AES would be used for this purpose
- Data security including local storage and data over the network
 - AES encryption/decryption
- MPC is used for memory protection
- Secure JTAG debugging
 - Supported by M33

7. HiFi 5 DSP Core

A Tensilica HiFi 5 DSP plus memory subsystem is dedicated for customer and third-party algorithms.

7.1. Hardware

- 32-bit fixed-point and floating-point
- 500 MHz maximum DSP clock
- 1 MB local RAM, plus additional 1.5 MB shared RAM (total 2.5 MB available memory)
- TCM
 - 64 KB instruction
 - 128 KB data
- Cache
 - 32 KB instruction
 - 32 KB data
 - Prefetch
- On-chip debug (OCD)/Tensilica TRAX
- Configurable interrupt controller external to the core

7.2. Software

100% open platform running on FreeRTOS (or customer's preferred embedded RTOS).

- Software APIs provided to interact with SoC resources:
 - Control interface to applications running on other processors
 - Control interface for hardware configuration
- Flexible audio routing to and from other processing points within the system

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8. Synaptics DSP Subsystem

8.1. Overview

The SR80 Series DSP Subsystem is designed with the following high-level objectives in mind:

- Low power consumption
- Low latency data paths for ANC and ambient fusion
- High-compute processing in low latency data paths
- Flexible memory architecture to support a wide variety of use-cases
- Flexible audio processing platform allows for feature addition in the future
- Customer-configurable DSP

There are three Synaptics DSP cores in this subsystem:

- Two CAPE 2 DSPs
- One low latency engine (LLE – vector processor based on CAPE architecture)

The DSP system memory architecture was designed as the best possible compromise between low-power, flexibility, and cost. There are small pools of memory banks dedicated to each processor. The dedicated pools of standard SRAM (64-bit wide SRAM) banks can be accessed via each processors program, data X, or data Y ports. A dedicated pool of vector SRAM (384-bit wide SRAM) banks can be accessed on vector data X and vector data Y ports. All DSP subsystem SRAM can be accessed via AHB.

Shared local memory is 0.5 MB. This memory is also accessible by the HiFi 5 DSP core.

Mechanisms available for passing data throughout the system:

- System memory
- DMA
- Low latency data path used for low latency single-sample audio routing between DSPs and other hardware elements

8.2. DSP Subsystem Functions

Synaptics DSP engines service these high-level algorithms and tasks:

- ANC: Active noise cancellation
- Ambient Fusion: Processing of sounds external to the ear and subsequently mixed into headphone playback.
- SRC: A variety of concurrent synchronous (SRC) and asynchronous (ASRC) audio sample rate conversions within the 8 kHz to 192 kHz range.
- Playback processing: processing of music playback or voice Rx signal, with modules such as EQ and dynamic range control (DRC).
- Transmit:
 - Tx processing: Primary processing on the transmit communication path such as analysis, noise reduction, acoustic echo cancellation (AEC), and synthesis.
 - Tx post-processing: Processing of the Tx signal in the final stages before passing the signal to the transmit device such as EQ and DRC.
- Voice Wake and Voice Command
 - Keyword detection

8.2.1. Hybrid ANC

The SR80 Series provides high quality low-power active noise cancellation (ANC), with ultra-low latency. The full processing path is all digital, providing for consistent performance, ease of tuning, and algorithm upgradability/configurability.

The ANC algorithm makes use of both an external microphone (feedforward cancellation) and an internal microphone (feedback cancellation) to achieve optimal results. The algorithm can operate in either a static non-adaptive mode or in adaptive mode. Audio intended to be heard by the user, such as music playback or the receiving channel in a voice call, is not altered by the ANC algorithm, nor does the audio degrade the ANC performance.

Ambient sounds can be seamlessly mixed into the headphone speaker output along with anti-noise produced by ANC, enabling users to flexibly manage which ambient sounds they hear and which are filtered out.

8.2.2. ENC

Noise suppression algorithms significantly improve the quality of voice pickup in noisy environments and allow for decent quality voice pickup either with no boom or a short boom.

The noise suppression algorithms employ advanced signal processing techniques to single out the user's voice and suppress all other sounds picked up by the microphones.

Noise reduction for all major electronic devices is available.

8.2.3. Sample Rate Conversion

Sample rate conversion (SRC) supports common rates from 8 kHz to 192 kHz. SRCs are typically used to convert between processing rates for different algorithms (software modules) running on either Synaptics DSPs or the Tensilica HiFi 5 DSP. Low-MIPS/high-quality SRCs are used for integer ration conversions. Conversions between 44.1 kHz and 48 kHz are provided.

Both synchronous and asynchronous sample rate converters are available.

8.2.4. Playback Processing

Equalization (EQ) and dynamic range control (DRC) can be applied to the playback audio signal. The EQ is a multiband IIR design allowing for flexible control over frequency response.

Playback processing also includes a simple tone generator and file playback, both mixed into the playback stream before DRC is applied. Tones and file playback can be used for various user audio notifications.

8.3. CAPE 2

The CAPE 2 DSP is the latest generation of Synaptics audio DSPs, designed specifically for high-quality, low-power audio processing.

- Flexible and easy to use memory architecture
- Dual 32-bit IEEE floating point MACs
- Dual 32-bit fixed-point MACs
- Quad 16-bit fixed-point MACs
- Eight 8-bit fixed-point MACs
- Dual 64-bit data memory buses
- Extensive arithmetic instruction set
- Complex arithmetic
- Program cache

8.4. LLE

The low latency engine (LLE) is a variant of the CAPE 2 processor with vector processing extensions. It is optimized for low-latency/high-compute audio processing applications such as ANC.

- 16 24-bit fixed-point MACs
- Dual 384-bit vector memory data buses
- Extensive arithmetic instruction set
- Complex arithmetic
- Program cache

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9. Synaptics NPU

Synaptics easy-to-use and highly optimized AI Neural Network acceleration processor working seamlessly with HiFi5 DSP supporting TensorFlow Lite for Microcontrollers.

- Easy to use and flexible NN processor with provided Synaptics SynAI optimized compiler supporting offline weight compression and easy to use and debug simulator.
- Supporting most computation intensive TFLM NN operators for advanced AI voice and audio applications.
- 32 MACs (8 x 8-bit and 16 x 8-bit)
- On-the-fly NN model weight decompressor
- 500 MHz maximum core speed
- 75% average MAC usage
- 32@500 MHz GOPS
- 32 kB local buffer and flexible allocation of low-latency shared system memory

The Synaptics NPU is supported with easy-to-use tools, including compiler that generates optimized models, enabling the efficient assignment of operators, as well as a simulator for offline development, and an NPU Profiler for in-depth analysis.

The Synaptics NPU supports NN models of various architectures, and upgradeable operators.

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10. Memories

The SR80 Series contains many different internal memory blocks. Memory blocks may be local or external to a processor and may be available or restricted based on memory protection settings. Internal memory can be split into three memory types: OTP, RAM, and ROM.

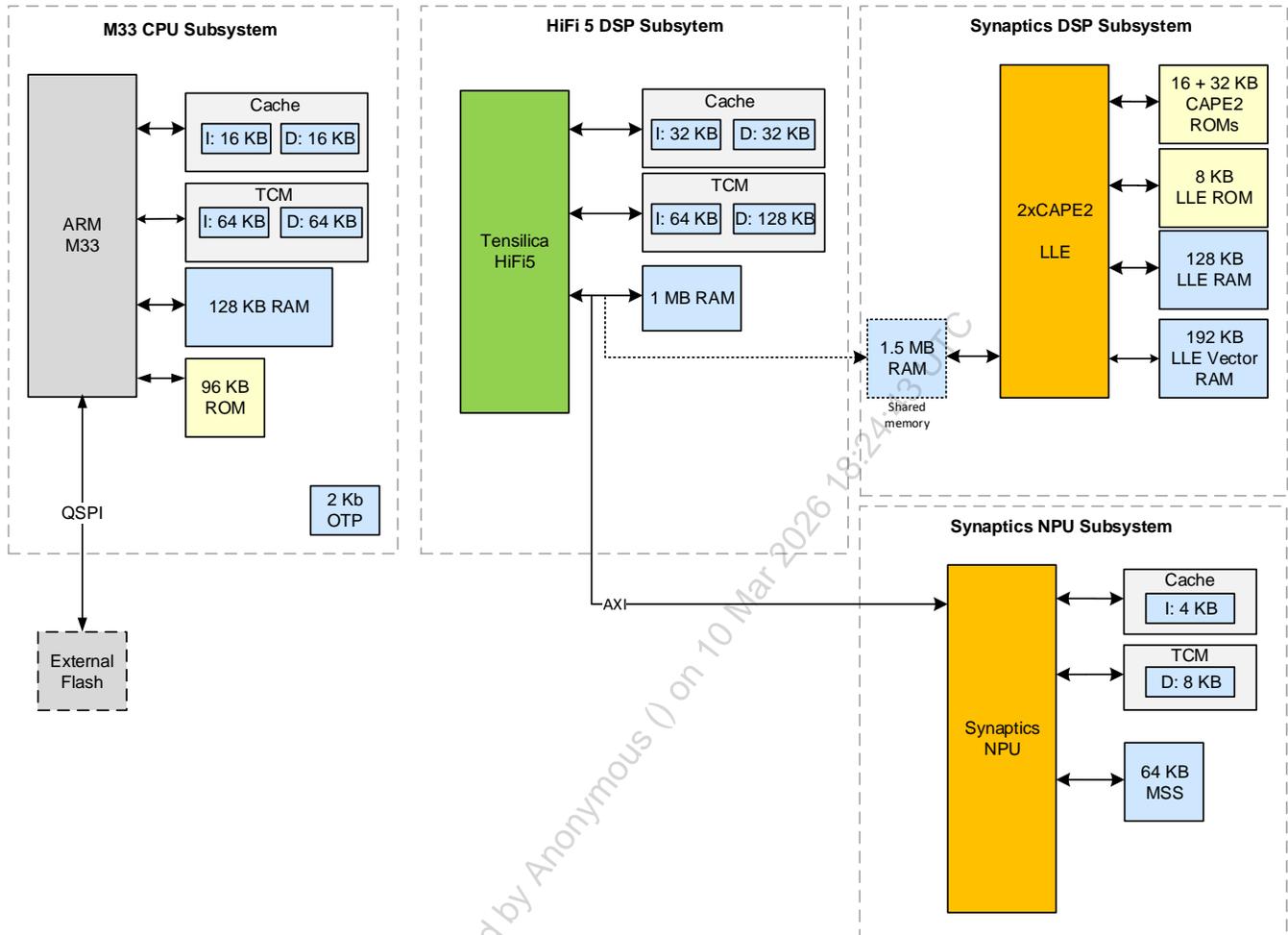


Figure 16. SR80 Series Memories

10.1. OTP

A 2 Kbit one-time programmable (OTP) memory is used for persistent storage of production test calibration values and security keys.

10.2. RAM

The total RAM memory of approximately 3.5 MB is divided into:

- System memories
- Local memories

Within memory, there are vector memory banks and standard (scalar) memory banks. There are five DSPs (CAPE 2 A, CAPE 2 B, LLE, NPU, and HiFi 5) and a CPU (Arm Cortex-M33) sharing these memory banks.

- Standard memory:
 - Each processor has several banks of its own local standard memory
 - Local and cache memories are accessible by other processors when not in use
 - In addition, there is a pool of sharable banks that every DSP can use
 - Standard memory is byte-addressed in 64-bit lines
 - All DSPs use standard memory for X data, Y data, and P program
- Vector memory:
 - A separate subsystem used by the LLE

10.2.1. Local (dedicated) Memory

Each processor has local tightly coupled SRAM usable for either program or data. The Synaptics DSP Subsystem has a total of 1.5 MB of local shared SRAM, and additional 128 KB RAM for LLE.

10.2.2. System (scalar) Memory

Each processor can access system memory. Accesses to system memory are arbitrated with AHB as the highest priority. System memory is 64 bits wide.

10.2.3. Vector Memory

- 192 KB RAM
- Used by LLE

10.3. ROM

The Arm CPU and Synaptics DSP subsystems have ROM for boot code and other static information.

11. Clock and Reset

11.1. Clock Sources

The SR80 Series uses the following clocks and time-base sources:

- Crystal/MCLK
- Standard RC oscillator (STD_OSC)
- Low-power RC oscillator (LP_OSC)
- I2S0 BCLK
- I2S1 BCLK
- I2S2 BCLK
- USB start-of-frame (SOF)

11.2. Crystal/MCLK

For USB-only applications, the SR80 Series requires the use of a 24 MHz crystal, with the following specifications:

- Frequency: 24 MHz
- Frequency tolerance: 400 ppm
- Peak jitter: +/- 100 psec
- Duty cycle: 40/60 max.

For other applications, an external clock source at MCLK can be used (such as a DECT or Bluetooth device), applied at the XTAL_I/MASTER_CLK pin.

The supported MCLK frequency range is 512 kHz to 25 MHz.

11.3. STD and LP Oscillators

The integrated standard RC oscillator (STD_OSC) in the SR80 Series generates a clock in the 16–32 MHz range. This clock is suitable as the clock source for audio use cases (PLL reference clock, ADC/DAC clock) when an external clock source is not available.

The STD_OSC oscillator is factory calibrated during production testing to ~1% frequency accuracy, with ~3% frequency variation across temperature.

The STD_OSC oscillator can also be adjusted to the same frequency as XTAL_I/MASTER_CLK, which can minimize frequency jumps when switching from external to internal clock source.

The low-power RC oscillator (LP_OSC) in the SR80 Series generates a 200 kHz internal clock. It is intended to provide a low-power clock for non-audio, low-power use cases (for example, standby or sleep).

The LP_OSC oscillator is factory calibrated to ~5% frequency accuracy.

11.4. PLLs

Two PLLs are integrated into the SR80 Series: one is the integer PLL, and the other is the fractional PLL.

Both PLLs use the following as reference clock inputs:

- I2S0, I2S1, I2S2 BCLK inputs
- XTAL_I/MASTER_CLK input
- STD_OSC internal clock

The I2S BCLK input frequency range supported is 512 kHz to 24.576 MHz.

The outputs of the PLLs are used to provide the needed clocks for the various sub-systems (audio, M33, HiFi 5, NPU, and so on), with appropriate clock dividers as needed.

11.5. PLL Reference Management

Inputs to the PLLs are managed by the PLL Reference Management block.

11.6. Reference On/Off Detection

Reference On/Off Detection block generates interrupts when the external reference is turned off or on.

Each PLL reference has a dedicated Reference On/Off Detection mechanism.

Ideally, software should be aware of when the external reference will be turned off, to manage the transition accordingly and to minimize impact on audio.

When in Automatic mode, the reference selection can be switched to internal reference when the external reference is detected to be off. This ensures that there is no loss of clock but does not necessarily ensure clean audio transitions.

11.7. USB Start-of-Frame (SOF)

The USB sub-system provides a start-of-frame time reference (1 kHz for FS, 8 kHz for HS) that is used for ASRC and clock measurements.

11.8. Clock Measurement

The clocking scheme in the SR80 Series includes a Clock Measurement block that measures the phase/frequency of various clocks to allow reference tracking and controlling the PLL output frequency accordingly.

11.9. Reset

The SR80 Series has a dedicated active-low chip reset pin (RESETN). A logic *HIGH* at this pin brings the device out of reset, provided the internal power management has completed power-on enable sequence (with PMU_EN signal).

The RESETN pin has an integrated pull-up resistor; an external resistor and capacitor can be used for more precise reset timing.

12. I2S/TDM Digital Audio

The SR80 Series device integrates three I2S/TDM digital audio interfaces, enabling digital audio connectivity in digital systems such as Bluetooth and DECT.

All three I2S interfaces can operate as host (transmit clocks) or target (receive clocks).

In target mode operation, the I2S BCLK inputs can be used as inputs to the internal clock generation blocks.

I2S0 and I2S1 support 6 RX channels, 6 TX channels.

I2S2 supports 4 RX channels, 4 TX channels.

The I2S interface block supports the following modes and configurations on the I2S/PCM bus:

- Word lengths of 8, 16, 24 bits
- Supports Philips (I²S)/left justified/right justified/PCM/TDM modes.
- Supports host and target mode.
- Supports sharing of the WS signal between transmitter and receiver.
- Programmable length of WS signal in host mode.
- Data may be driven on falling or rising edge of BCLK.
- Programmable channel order: 'left then right' or 'right then left'
- BCLK frequency range: 512 kHz to 24.576 MHz

12.1. I2S Mode Timing

The I2S timing uses WS (LRCK) to define whether the data is being transmitted for the left channel or for right channel. The WS is low for the left channel and high for the right channel. A WS polarity control bit is provided to allow either high or low to represent the left channel. The default setting of the polarity control is 0, which means low WS = left channel. The WS need not be symmetrical.

A system clock (BCLK) running at a minimum of $2 \cdot (\text{sample width}) \cdot (\text{sample frequency})$ is used to clock in the data. There is a delay of one clock bit from the time the WS signal changes state to the first data bit on the data line. The data is written MSB first and is valid on the rising edge of the bit clock. When the programmed sample width is taken, any remaining bits are ignored.

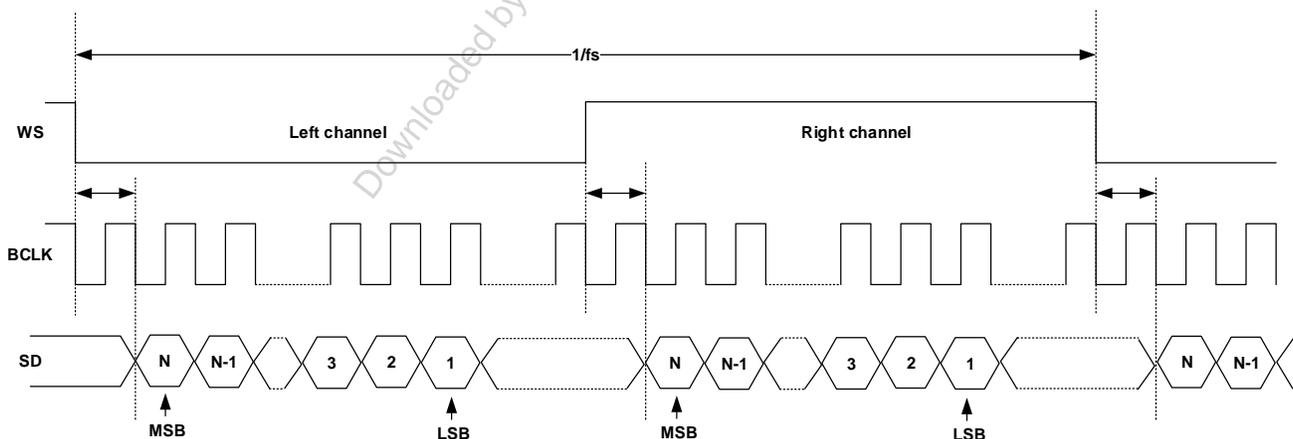


Figure 17. I2S timing diagram – width of LRCLK frame is wider than $2N$ bits ($N = 8, 16, 24, \text{ or } 32$)

12.2. Left-justified Mode Timing

Left-justified timing uses the WS clock to define when the data being transmitted is for the left channel and when it is for the right channel. The WS is high for the left channel and low for the right channel (requires the WS polarity control bit to be set to 1). A bit clock running at a minimum of $2 \cdot (\text{sample width}) \cdot (\text{sample frequency})$ is used to clock the data.

The first data bit appears on the data lines at the same time WS toggles. The data is written MSB first and is valid on the rising edge of the bit clock. When the programmed sample width is taken, any remaining bits are ignored. If the WS toggles before the full word length is read, the remaining bits are zeroed.

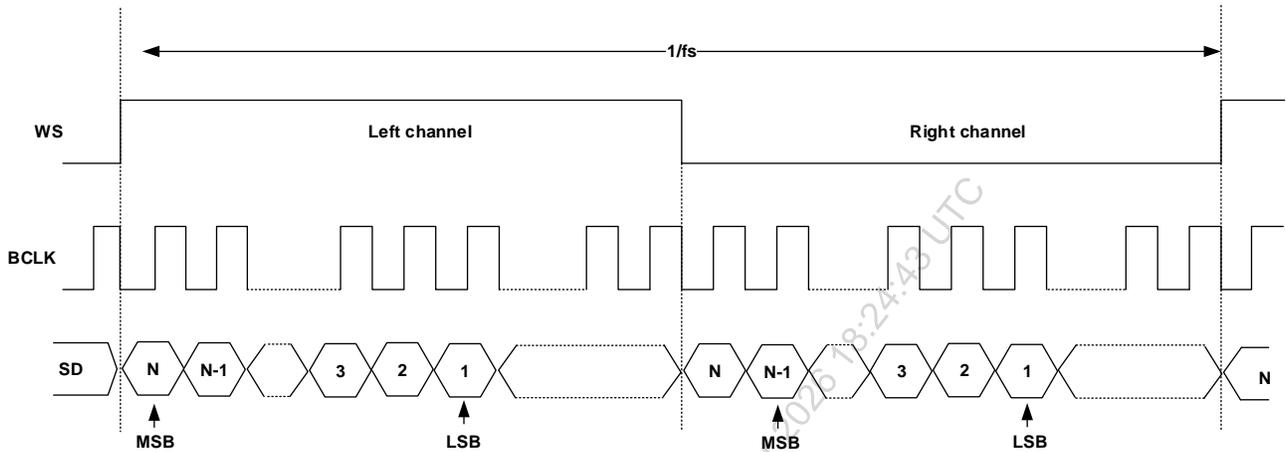


Figure 18. Left-justified timing diagram

12.3. Right-justified Mode Timing

Right-justified timing uses the WS clock to define whether the data is being transmitted for the left channel or for the right channel. The WS is high for the left channel and low for the right channel (requires the WS polarity control bit to be set to 1). A bit clock running at a minimum of $2 \cdot (\text{sample width}) \cdot (\text{sample frequency})$ is used to clock the data.

Data is captured in a 32-bit shift register until the WS toggles. When the WS toggles, the last 32, 24, 16, or 8 bits are transferred to the channel indicated by the previous state of WS. In right-justified mode, the LSB of data is always clocked by the last bit clock before the WS transitions. The data is written MSB first and is valid on the rising edge of the bit clock. All leading bits are ignored.

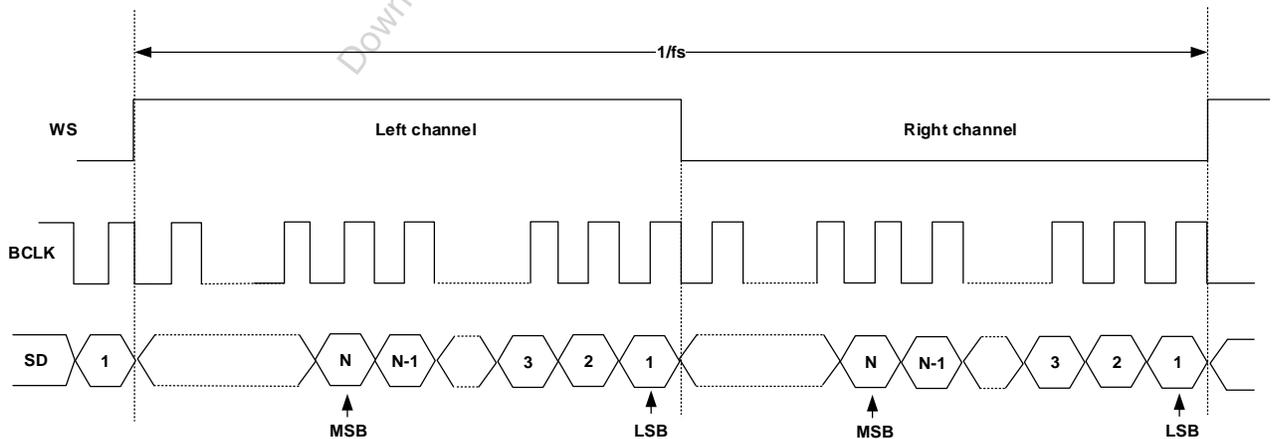


Figure 19. Right-justified timing diagram

12.4. PCM Short-frame Mode Timing

In short-frame mode, the falling edge of PCM_SYNC indicates the start of the PCM word. The PCM_SYNC is one clock long. Data is driven out on the rising edge of PCM_CLK after the PCM_SYNC pulse.

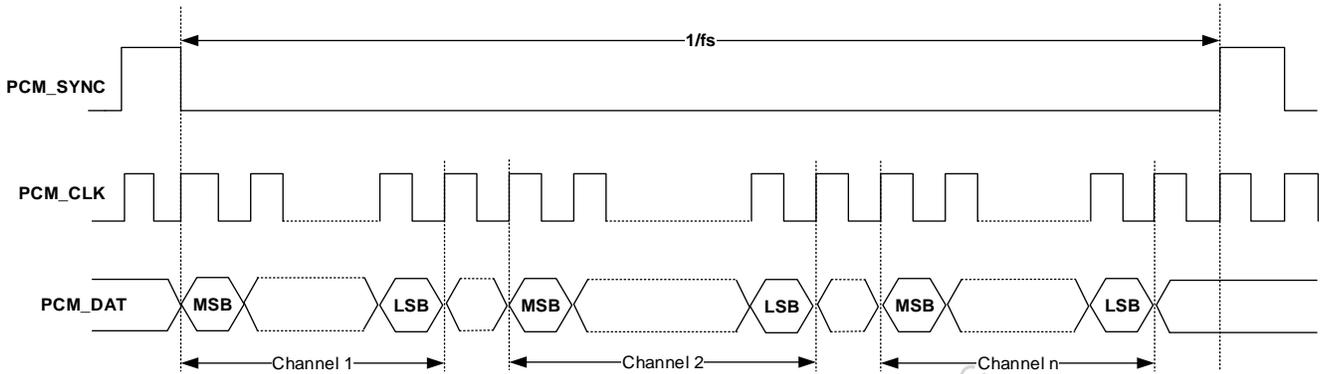


Figure 20. PCM short-frame timing diagram

12.5. PCM Long-frame Mode Timing

In long-frame mode, the rising edge of PCM_SYNC indicates the start of the PCM word. The PCM_SYNC is at least two clocks long. Data is driven out on the rising edge of PCM_CLK coincident with the rising edge of PCM_SYNC.

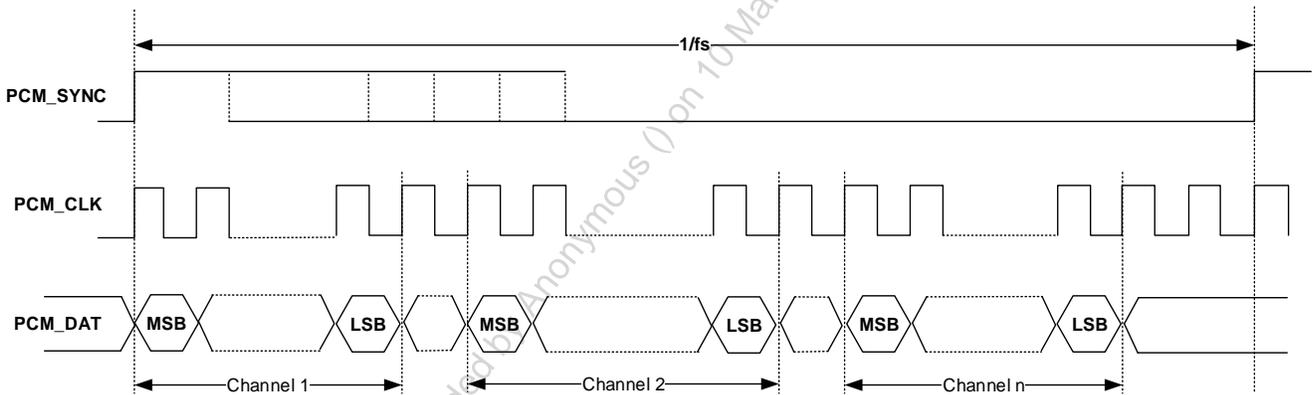


Figure 21. PCM long-frame timing diagram

12.6. PCM Multi-slot Mode Timing

In multi-slot mode, PCM_SYNC can be either long or short. Four words of data can be sent or received. The start of the PCM word position is determined by the length of the sync pulse. Slots are determined by counting data width clocks (8, 16, 24 or 32) from the first PCM word. The PCM mode supports up to four channels. While transmitting, the channel selection should always be from Channel_1.

13. Digital Mic PDM

The SR80 Series device supports up to 8 digital microphones, using four DMIC interfaces. Each interface consists of one output clock and one input data and is typically connected to a pair of digital microphones.

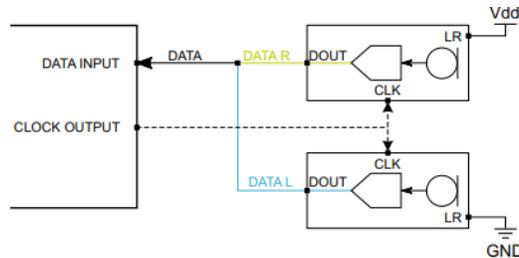


Figure 22. DMIC interface usage

The DMIC clock frequency supported is from 256 kHz to 6.144 MHz.

DMIC interface functionality:

- During the rising edge of the clock the digital microphone with select=0 is active, and the other digital microphone is in high-Z.
- During the falling edge of the clock the digital microphone with select=1 is active, and the other digital microphone is in high-Z.

14. I2C/I3C

The SR80 Series includes the following I²C/I3C interfaces:

- I2CO Host
- I2C1 Host
- I2C Target
- I3C Host/Target

The I²C target interface and the I3C interface share the same pins so only one mode of operation can be used.

14.1. I2CO and I2C1 Host

The I²CO and I²C1 host in the SR80 Series works at up to 400 kHz clock rate (Fast mode) and includes support for clock stretching to pause I²C communication as needed.

14.2. I2C Target

The I²C target interface in SR80 Series is used for control from an I²C host.

The I²C target works at up to 1 MHz clock rate (Fast mode Plus) and implements clock stretching to pause I²C communication as needed.

14.3. I3C Host/Target

The SR80 Series integrates a 2-wire I3C host/target interface for fast boot and control.

At initial boot-up, the interface starts as I²C target and can be switched to I3C during the I²C boot sequence, via a command. After I3C mode is verified, I3C fast boot can be used.

I3C features:

- Various operational modes
 - I3C controller (host or target)
 - I²C target
 - Data transfer to legacy I²C target devices
- Dynamic Address Assignment (DAA)
- Hot-Join feature support
- In-band interrupts (IBI)
- Data rates: FM, FM+, SDR, HDR-DDR
- CRC/parity generation and validation
- Support for broadcast and directed CCC transfers

Note that the SR80 Series I3C interface shares the same pins as the I²C target.

15. 1-Wire PLC Communication

The SR80 Series features a 1-wire interface for communication between two devices, intended for communication using the 5-V charging wire (and GND) while the earbud is inside the charging case. This is also sometimes referred to as power-line communication (PLC). Note that while in 1-wire communication mode, charging is paused until communication mode is exited.

The communication is half-duplex, using a UART-like, software-implemented protocol.

In the transmit direction, one of the PWM LED outputs (*LED_SINKO*) is used as this is high-voltage tolerant pin. In the receive direction, *UART2_RX* is used.

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16. Serial Peripheral Interface (SPI) Host

The SR80 Series device features two SPI host interfaces. It supports the following modes:

- Mode 0: CPOL = 0, CPHA = 0
- Mode 3: CPOL = 1, CPHA = 1

The SPI host interface is intended to interface with SPI target devices, such as a serial flash or sensors. The SPI block consists of an 8x32-bit Tx and Rx first in, first out (FIFO) with several registers for control and status. The SPI accommodates various controls for clock polarity, a phase shift of the clock, and target select polarity.

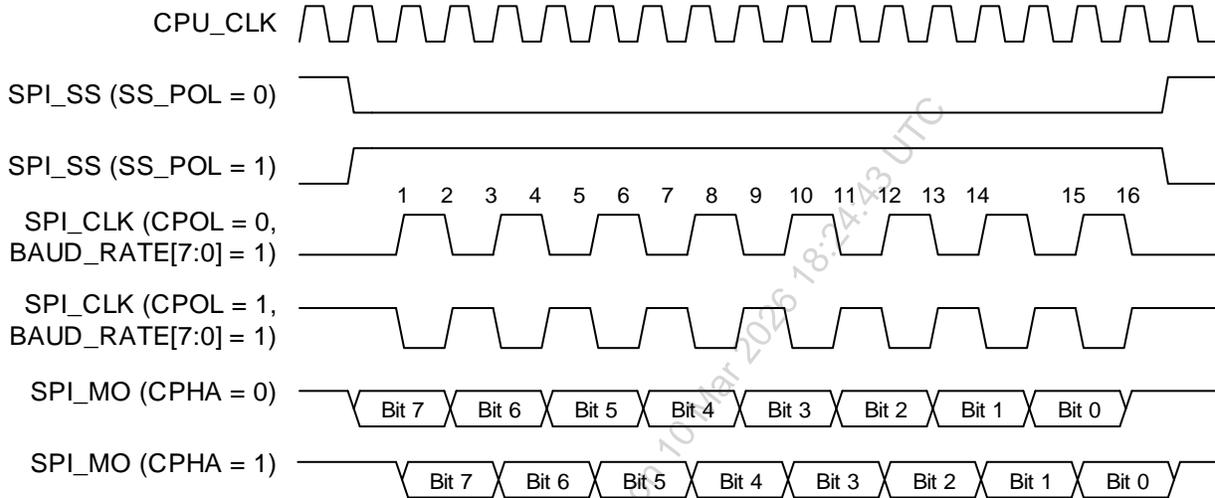


Figure 23. SPI Tx timing

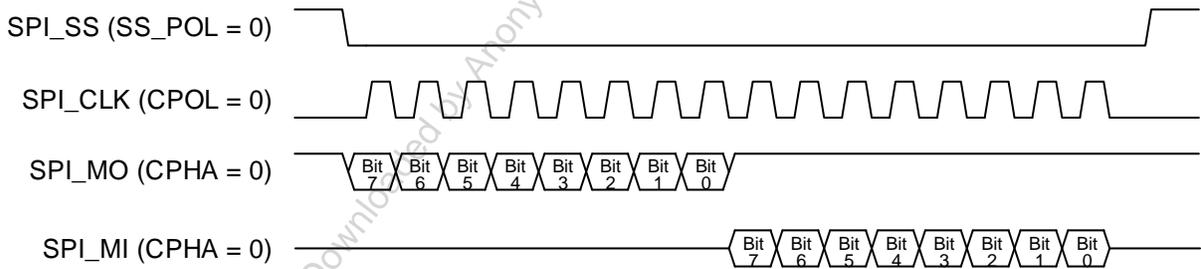


Figure 24. SPI Tx and Rx timing

17. QSPI Host/SPI Target

The SR80 Series features a QSPI host interface which provides a mechanism for accessing external devices (for example, flash memory devices) through a serial bus referred to as a serial peripheral interface (SPI).

The QSPI host interface can also be re-purposed to be a SPI target interface, which can eliminate the need for a dedicated external flash device by booting from an external SPI host, and for control of the SR80 Series.

17.1. QSPI Host

A standard SPI bus utilizes a chip select, a clock, a serial data output and a serial data input signal. The QSPI host interface supports the standard SPI bus, as well as variations of the SPI bus that provide for increased throughput. These variations include the use of multiple (2 or 4) data signals, and the ability to use data signals in a bi-directional fashion.

The QSPI host interface supports clock speeds up to 120 MHz.

The QSPI host interface supports “on-the-fly” AES-256 decryption of flash data for increased security.

The QSPI host module provides four mechanisms for initiating transfers to/from external memory.

- Programmed DMA
 - The module includes an internal DMA engine, which can be programmed to transfer data between on-chip memory and externally attached device(s). This mechanism makes use of an AHB host interface on the module to access on-chip memory through an on-chip AHB bus. This mechanism may be used to transfer data in either direction (from internal memory to external device or from external device to internal memory).
- Pre-load DMA
 - The same internal DMA engine used for programmed DMA is also used to implement a pre-load function. The pre-load function executes automatically after a power-on reset and transfers data from an external device (which should typically be a memory device) to on-chip memory. A data structure stored at the head of the external memory is used to control which data is moved from the external memory, and to what locations in internal memory it is moved.
- Execute in place (XIP)
 - The module includes an AHB-Lite target interface. The module responds to read accesses within a programmable range of AHB address space by retrieving data from the externally attached memory and delivering it through the AHB-Lite interface. This mechanism can be used only to transfer data from an externally attached memory to the internal AHB bus. This mechanism is intended to support software execution directly from the externally attached memory. For optimal performance, a cache (external to the QSPI module) is recommended.

Note that some flash memories support a special XIP mode, which allows the usual command/instruction byte to be skipped during XIP transfers. This mode is supported by the XIP transfer initiation mechanism interface. However, the other transfer initiation mechanisms (DMA or register-based) may not function correctly while an attached memory device is configured in the special XIP mode.

- Register-based transfers
 - The module includes a register-based interface for initiating individual transfers to/from the external memory.

The QSPI interface supports four modes of operation, MODE0–MODE3. The default mode of operation is in MODE3.

- MODE0
 - Figure 25 illustrates mode 0. In mode 0, the clock idle state is low, and the first data bit of a transfer is captured (by either the host or target) on the first clock transition of the transfer.

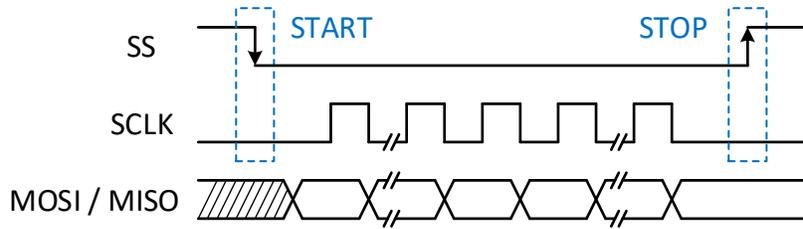


Figure 25. MODE0 timing

- MODE1
 - Figure 26 illustrates MODE1. In MODE1, the clock idle state is low, and the first data bit of a transfer is captured (by either the host or target) on the second clock transition of the transfer.

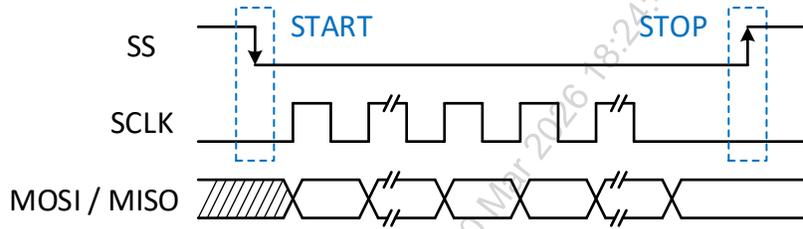


Figure 26. MODE1 timing

- MODE2
 - Figure 27 illustrates MODE2. In MODE2, the clock idle state is high, and the first data bit of a transfer is captured (by either the host or target) on the first clock transition of the transfer.

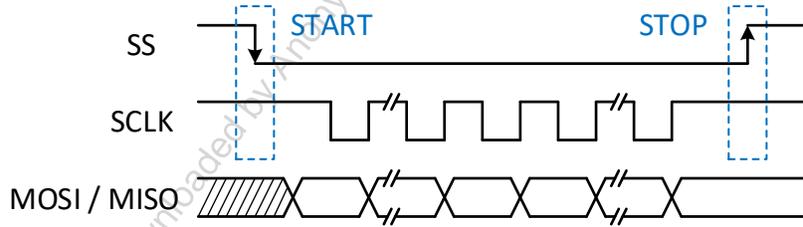


Figure 27. MODE2 timing

- MODE3
 - Figure 28 illustrates MODE3. In MODE3, the clock idle state is high, and the first data bit of a transfer is captured (by either the host or target) on the first clock transition of the transfer.

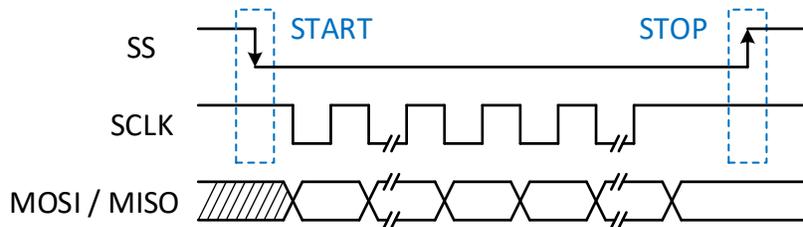


Figure 28. MODE3 timing

Basic, fast, dual output, dual I/O, and quad I/O instructions are supported. In addition, these transfer characteristics are configurable:

- Instruction value
- Address length (bytes)
- Dummy cycles (clock cycles)
- Data length (bytes)
- Single, dual, or quad address
- Single, dual, or quad data

17.2. SPI Target

The SPI target interface can be used by a SPI host for booting and control, enabling the sharing of a single SPI flash memory device.

The following modes are supported:

- Mode 0: CPOL = 0, CPHA = 0
- Mode 3: CPOL = 1, CPHA = 1

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18. Flow Control UART

The SR80 Series features three universal asynchronous receiver transmitter with flow control (FC UART), compatible with a standard 16550 UART. Most register bit functions, bit locations within registers, and register base offsets are the same as in a standard 16550 UART. Compared to the 16550, the FC UART operates at a higher clock frequency and provides a flow control mechanism.

- Supports 7-bit data or 8-bit data
- Supports 1 or 2 bit stop bit
- Supports 1-bit odd or even parity
- Supports DMA or CPU channel accessing RX/TX FIFO
- Supports IRDA mode
- The baud rate can be configured by firmware
- Supports flow control

Baud rate and other configuration options are selected using the application framework managed by the CPU subsystem. Most common rates are supported, up to 3 Mbps.

The UART interface can be used for debug (for example, print messages), and for control of other devices in the system.

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19. PWM LED Sinks

The SR80 Series integrates 3 high-voltage tolerant current sinks, used to directly drive external LEDs, without the need for external transistors. Each pin can sink up to 20 mA of current.

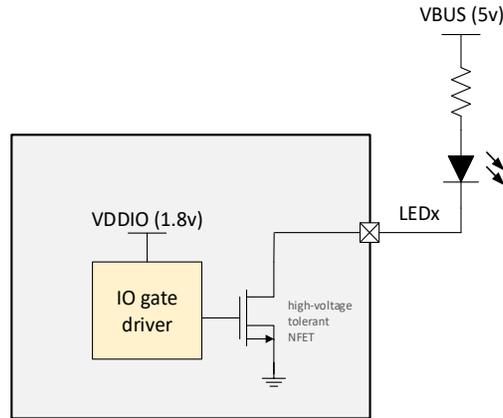


Figure 29. PWM led sink in SR80 Series

- Internal PWM logic is used to control the frequency, duty cycle, and blink rate.
- The maximum PWM frequency is 8 kHz. 8-bit dimming control is applied to the PWM signal.

The SR80 Series also includes a second set of PWM outputs, with standard I/O output capability. This second set is not high-voltage tolerant.

The two sets of PWM outputs are particularly useful for creating differentiated LED lighting for personal and enterprise devices.

20. MADC

The monitor ADC block can be used to detect button presses using a multi-button array and can also be used to measure system voltages. It uses a 10-bit SAR ADC.

The MADC will typically be operated in a power-cycled mode, where it is powered up briefly for a measurement, then powered down again, effectively consuming a small amount of power. The clock for the MADC is 200 kHz, provided by the low-power oscillator (LP_OSC).

The MADC can also be operated in a continuous mode, at ~15 ksps (200 kHz/13).

The measurement range for the MADC is 0 to 1.8 V, with absolute accuracy of +/- 15 mV maximum.

20.1. Multi-button Sensing

The multi-button sense input to the MADC offers the ability to connect multiple switches to the SR80 Series using a single pin. Typically, 1 to 4 buttons are connected using a resistive divider approach where each button momentarily connects a resistor to ground using different resistance values. Various topologies are possible.

The ultra-low-power button-sense input uses a comparator to detect button presses, which then wakes up the ADC for conversion. Firmware can then be used to implement button-press detection, including short and long presses.

20.2. Voltage Sensing

The MADC voltage sense input (at pin MICOP) can be used for general-purpose voltage monitoring.

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21. Power Management

Power is managed by the power management unit (PMU) and firmware running on the Arm M33 CPU.

21.1. Power Supplies

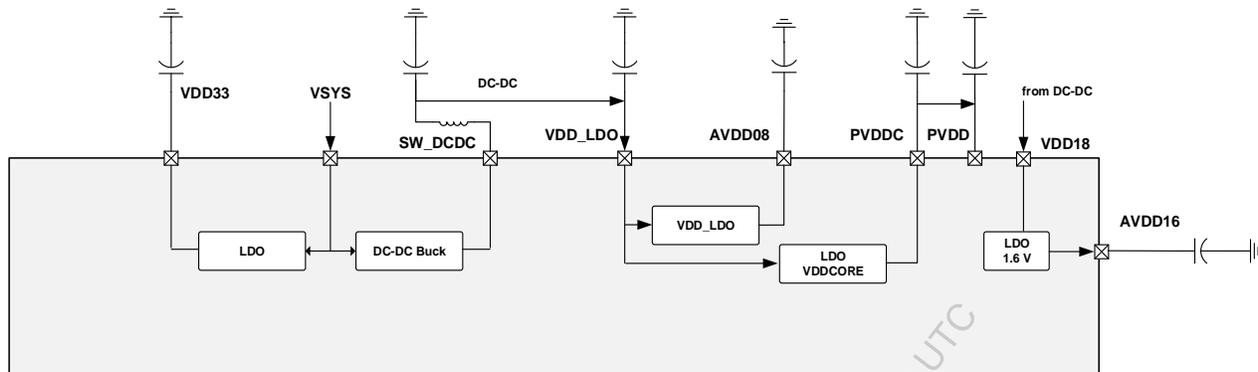


Figure 30. Power supplies in the SR80 Series device

Table 25. Power Rail Supply Summary

| Supply Name | Direction | Typical Voltage | Source / Dependency | Primary Usage / Notes |
|-------------|----------------------|--------------------------------|---|---|
| VSYS | Input | 3.7 V (battery) / 5.0 V (USB) | External battery or USB VBUS | Main system supply. Feeds DC-DC converter and 3.3-V LDO. |
| SW_DCDC | Output (switch node) | — | Internal DC-DC converter | Switching node. External inductor and capacitor generate regulated DC output (1.8 V or 1.0 V). |
| VDD33 | Output | 3.3 V (USB) / 2.85 V (battery) | Internal 3.3-V LDO from VSYS | Supplies USB PHY and microphone bias circuitry. Voltage programmable by use case. |
| VDD_LDO | Input | 1.8 V or 1.0 V | DC-DC regulated output | Input to core LDO and 0.8-V LDO. Also serves as DC-DC feedback node. |
| VDD18 | Input | 1.8 V | External (battery mode) or DC-DC (USB mode) | Input to 1.6-V LDO and I/O/HP driver supply. Must be externally supplied in 1.0-V DC-DC battery mode. |
| AVDD16 | Output | 1.6 V | Internal 1.6-V LDO from VDD18 | Supplies analog circuits (ADC, DAC, analog front-end). |
| AVDD08 | Output | 0.8 V | Internal 0.8-V LDO from VDD_LDO | Supplies PLL and high-frequency RC oscillator (HF-RCOSC). |
| PVDDC | Output | 0.8 V (programmable) | Internal core LDO from VDD_LDO | Core voltage rail for digital processing blocks. |
| PVDD | Input | Same as PVDDC | From PVDDC | Direct supply input to processing cores. |

21.2. Power Domains

There are three primary power domains in SR80 Series:

- M33 MCU power domain
- HiFi 5, NPU power domain
- DSP subsystem power domain (CAPE 2 and LLE cores)

21.3. Power States

Power states are primarily defined in firmware and managed by the CPU using the PMU and subsystem control registers. Power-on and wake from low power states are managed directly in hardware before transitioning to processor control. Power management uses a fine-grained approach to configure functional blocks as needed to perform the tasks at hand. This functionality not only allows adjusting the state of resources to handle the needed application tasks but also allows supporting applications with varying performance (quality/power) levels, for example low-power audio playback and high-quality audio playback.

21.4. Power Sequencing

21.4.1. *PMU_EN*

The Power Management Unit Enable (*PMU_EN*) pin is used to enable the internal power management. After VSYS is applied, the *PMU_EN* pin is **latched once** when it becomes logic *high*, enabling the device to come out of reset. Note that after this initial latching of *PMU_EN*, the pin simply becomes an input to the SR80 Series firmware. That is, a logic *low* at the pin does not automatically disable the internal power management.

PMU_EN can be connected directly to VSYS. Internally, the power management logic sequences the various power blocks to limit in-rush current.

21.4.2. *POR_BYPASS*

This pin should be tied to GND during normal use. It is intended for debug purposes, and for production testing of the SR80 Series.

22. System Overview and Application Diagrams

22.1. SoC Connectivity Overview

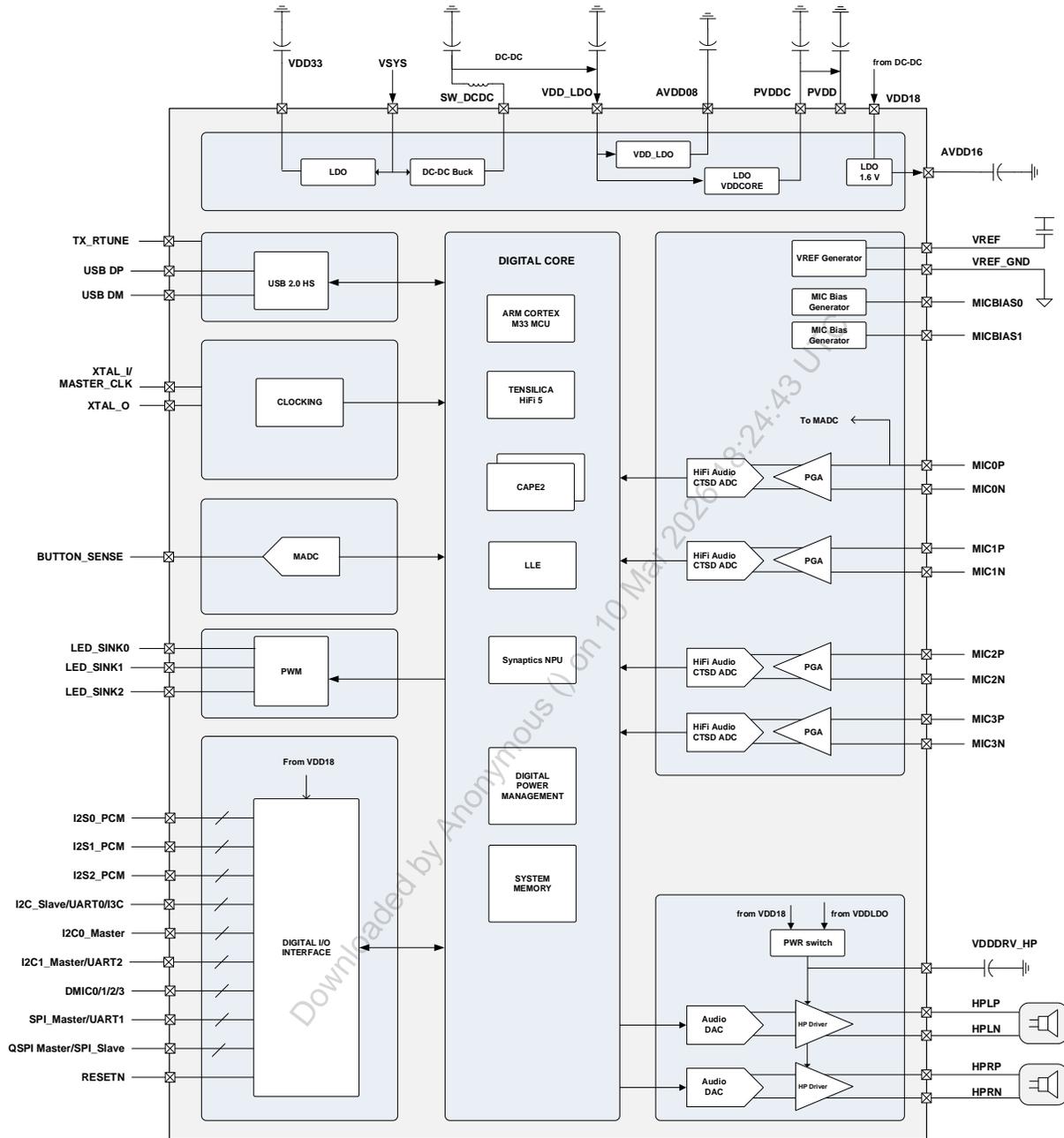


Figure 31. SoC Connectivity Overview

22.2. SR80 Series General Applications

- Advanced ANC headsets
- Office Unified Communications headsets
- Gaming headsets
- True-wireless stereo (TWS) headsets
- Wired USB headsets
- Speakerphones
- Bluetooth speakers
- USB microphones

22.3. Application Diagram with Two Stereo Speakers

Figure 32 illustrates the SR80 Series in electronic devices.

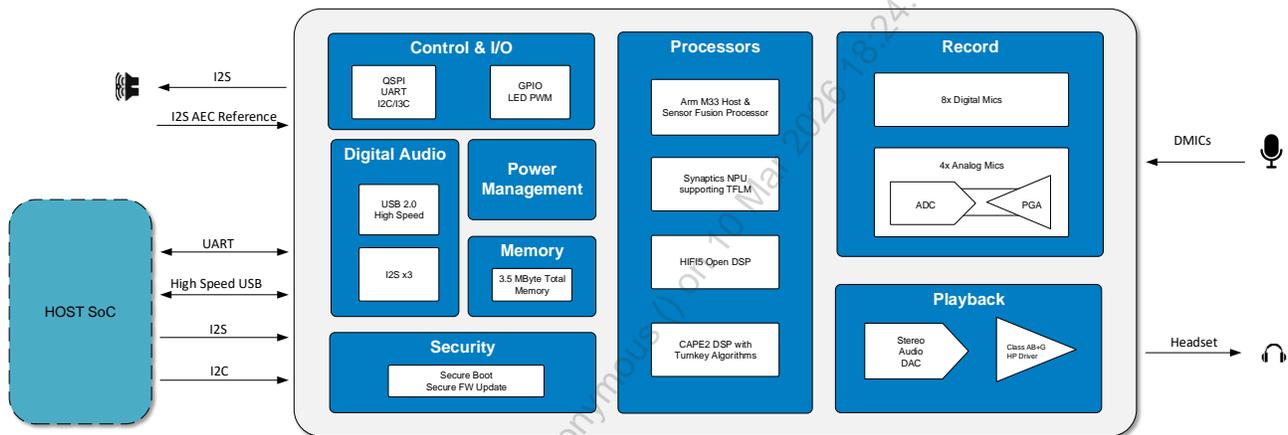


Figure 32. Electronic device application diagram with 2 stereo speakers (4 speakers optional)

22.4. Synaptics Audio Studio Application

The SR80 Series device is supported by the Synaptics Audio Studio application for tuning and configuring advanced audio processing and AI features.

- Graphical user interface
- Supports real time tuning, offline mode and firmware customizations
- Customizable memory layout
- Customizable DSP pipeline
- Easy to tune customer & third-party DSP modules

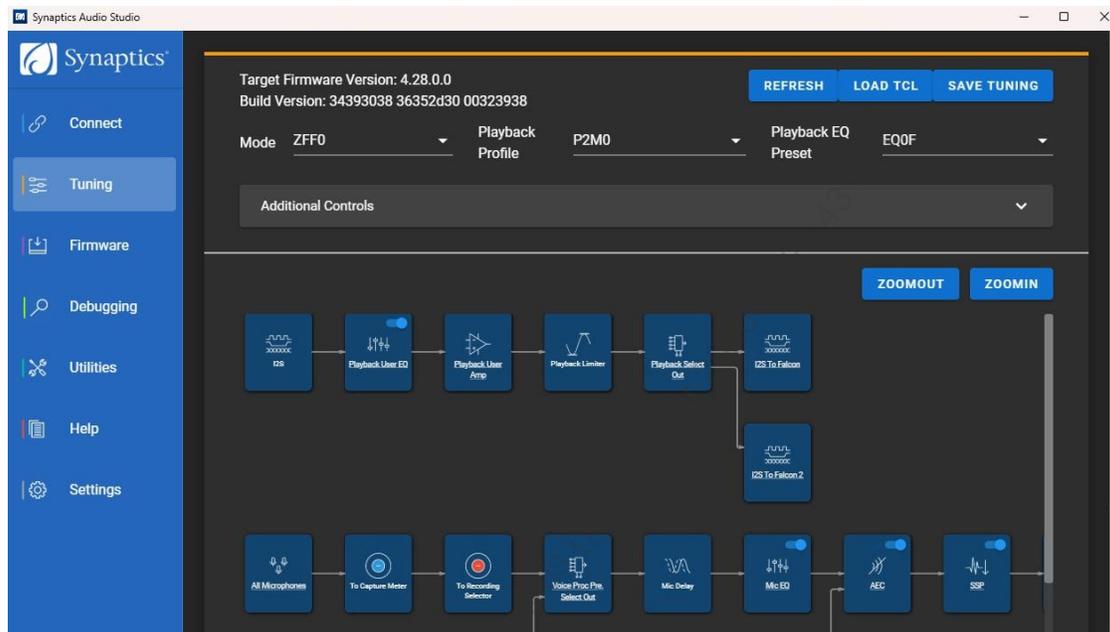


Figure 33. Synaptics Audio Studio application

Applications

- High-Performance Doorbells
- Intercoms
- Security Cameras & Panels
- Glass-Break & other "Listeners"

23. Mechanical and Package Information

23.1. Package Drawing and Dimensions

23.1.1. WLCSP-76 Package

The SR80 Series is available in 76-ball WLCSP package, with 0.35 mm ball pitch. Package dimensions are 3.08 mm x 3.20 mm. Full dimensions for the package are shown in Table 26.

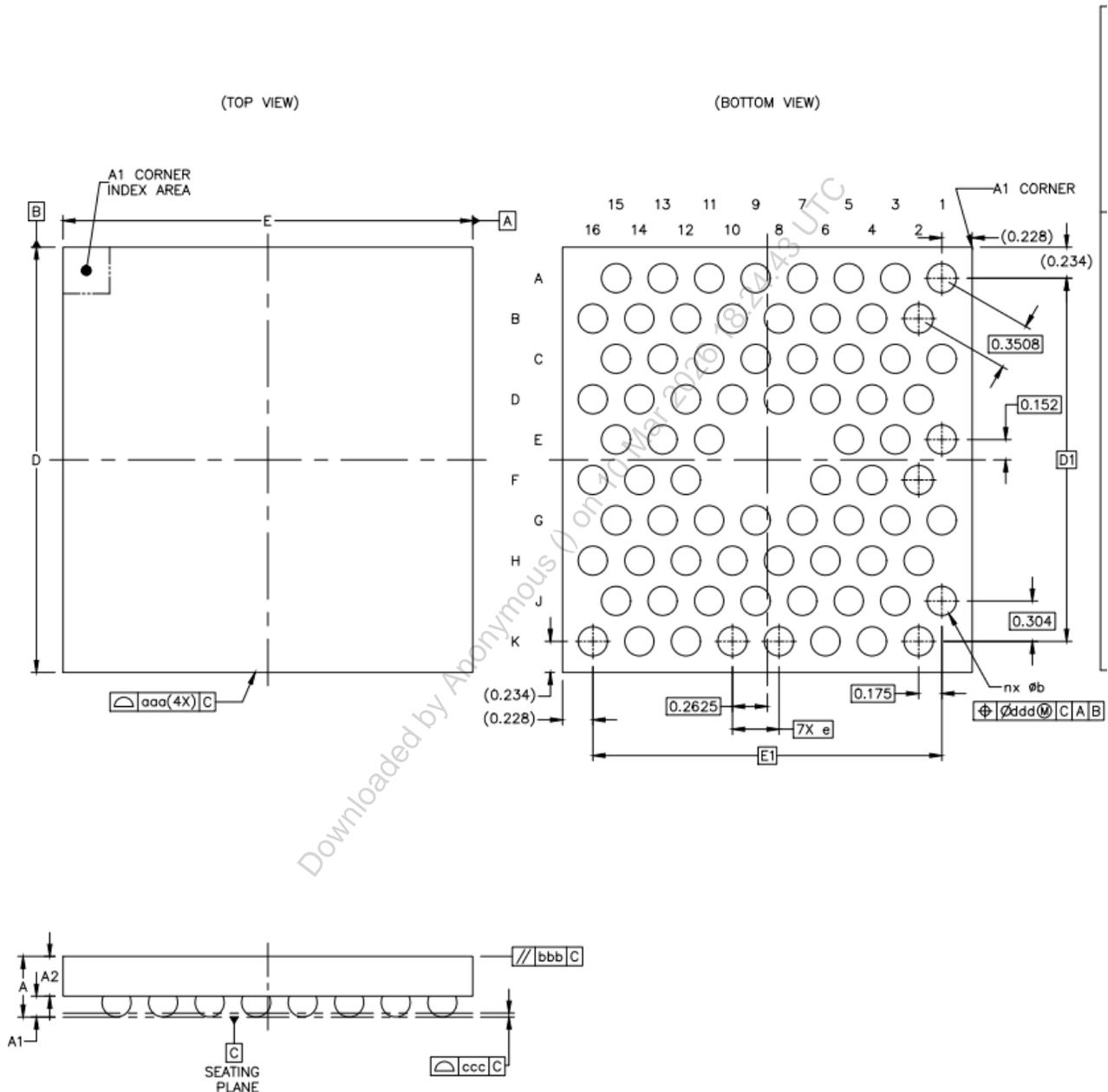


Figure 34. WLCSP package drawing for SR80 Series (units in mm)

Table 26. WLCSP Package Dimensions (in mm)

| | | Symbol | Common Dimensions | | |
|----------------------------|---|--------|-------------------|-------|-------|
| | | | Min. | Nom. | Max. |
| Total Thickness | | A | 0.432 | 0.457 | 0.482 |
| Standoff | | A1 | 0.137 | 0.157 | 0.177 |
| Wafer Thickness | | A2 | 0.285 | 0.3 | 0.315 |
| Film Thickness | | A3 | — | — | — |
| Body Size | X | E | 3.0809 | | |
| | Y | D | 3.2039 | | |
| Ball Bump Pitch | X | SE | — | BSC | |
| | Y | SD | — | BSC | |
| Edge Ball Center to Center | X | E1 | 2.625 | BSC | |
| | Y | D1 | 2.736 | BSC | |
| Pitch | | e | 0.35 | BSC | |
| Ball Diameter (Size) | | | 0.2 | | |
| Ball/Bump Width | | b | 0.19 | 0.22 | 0.25 |
| Ball/Bump Count | | n | 76 | | |
| Package Edge Tolerance | | aaa | 0.03 | | |
| Wafer Flatness | | bbb | 0.06 | | |
| Coplanarity | | ddd | 0.03 | | |
| Ball/Bump Offset (Package) | | eee | 0.015 | | |

Table 27. LGA Package Dimensions (in mm)

| | Symbol | Common Dimensions | | |
|-------------------------------|--------|-------------------|------|------|
| | | Min. | Nom. | Max. |
| Total Thickness | A | 0.57 | 0.62 | 0.7 |
| Substrate Thickness | A1 | 0.17 | | REF |
| Mold Thickness | A2 | 0.45 | | REF |
| Body Size | D | 6.5 | | BSC |
| | E | 6 | | BSC |
| Lead Width | W | 0.15 | 0.2 | 0.25 |
| Lead Length | L | 0.35 | 0.4 | 0.45 |
| Lead Pitch | e | 0.5 | | BSC |
| Lead Count | n | 76 | | |
| Edge Lead Center to Center | D1 | 4 | | BSC |
| | E1 | 5.5 | | BSC |
| Body Center to Contact Lead | SD | 0.25 | | BSC |
| | SE | 0.25 | | BSC |
| Pre-Solder | | — | — | — |
| Package Edge Tolerance | aaa | 0.1 | | |
| Mold Flatness | bbb | 0.2 | | |
| Coplanarity | ddd | 0.08 | | |

23.2. Package Thermals

Table 28. Thermal specifications

| Parameter | Symbol | Minimum | Typical | Maximum | Units |
|--|---------------|---------|---------|---------|-------|
| Thermal resistance ^{1,2} (four layer) | θ_{JA} | — | 37 | — | °C/W |
| Junction Operating Temperature | TJ | 0 | — | 125 | °C |
| Ambient Operating Temperature | TA | -10 | — | 85 | °C |

- For a given power dissipation, die temperature can be calculated as follows: $T_J = T_A + (\text{power dissipated} \cdot \theta_{JA})$.
- θ_{JA} uses a 4-layer 139.7 x 109.2 x 1.6 mm 2s2p test board.

Note: SR80 Series package is MSL1 (Moisture Sensitivity Level 1), indicating minimal sensitivity and no special handling requirements.

24. Part Numbering and Marking

24.1. Part Order Numbering

Table 29 lists the SR80 Series part numbers and their corresponding applications.

Table 29. SR80 Series Part Order Options

| Product Name | Part Number ¹ | Cores | Package Type |
|--------------|--------------------------|---|--------------|
| SR82 | SR82A1-LA0000-T000-R | MCU-CM33, DSP-CAPE2, LLE | LGA-76 |
| | SR82A1-WC0000-T000-R | | WLCSP-76 |
| SR85 | SR85A1-LA0000-T000-R | MCU-CM33, DSP-CAPE2, LLE and HiFi5 | LGA-76 |
| | SR85A1-LA0000-T000-R | | WLCSP-76 |
| SR88 | SR88A1-LA0000-T000-R | MCU-CM33, DSP-CAPE2, LLE, HiFi5 and NPU | LGA-76 |
| | SR88A1-WC0000-T000-R | | WLCSP-76 |

1. T=Tray, R=Tape & Reel

24.2. Package Marking

Figure 36 illustrates a sample package marking and Pin 1 location for an SR80 Series device.

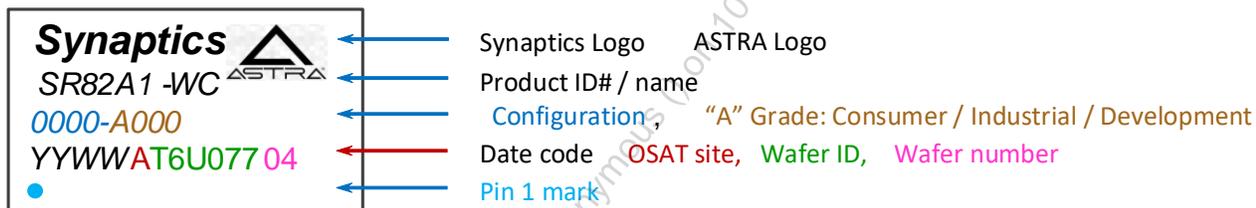


Figure 36. Package Marking and Pin 1 Location

24.3. Part Number Decoder

The following examples illustrate part numbers and their decoding process.

Example 1: SR82A1-LA0000-T000-R

Example 2: SR82A1-WC0000-T000-R

- SR82:** Chip name
- A1:** Revision name
- LA:** Package type: LGA76
- WA:** Package type: WLCSP76
- 0000:** Reserved for SW/Model/Algo Licensing
- A000:** Full feature engineering sample part
- V000:** Reserved for Automotive part
- H000:** Reserved for Industrial part
- T000:** Consumer part

Note: The same decoding method applies to all SR85/88 part numbers.

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25. References

Other related reference documents:

- I²C Specification: <https://www.nxp.com/docs/en/user-guide/UM10204.pdf>
- I3C Specification: <https://mipi.org/specifications/i3c-sensor-specification>

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26. Revision History

| Date Modified | Revision | Description |
|-------------------|----------|---------------------|
| February 25, 2026 | 1 | Initial release. |
| March 2, 2026 | A | Production release. |

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